

**DIGITAL OSCILLOSCOPE AND  
HIGH SPEED ANALOG CONVERSION  
CARD FOR THE DE2 DEVELOPMENT KIT**

**A Design Project Report**

**Presented to the Engineering Division of the Graduate School  
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Master of Engineering (Electrical)**

**by**

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# **Abstract**

**Master of Electrical Engineering Program**

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**Design Project Report**

**Project Title: Digital Oscilloscope and High Speed Analog Conversion Card for the DE2 Development Kit**

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**Abstract: This project covers the development of a hardware and software solution to high speed analog signal generation and capture on the DE2 FPGA board. The hardware portion consists of a daughter card for the DE2 that houses a digital-to-analog and an analog-to-digital converter and interfaces with the motherboard's Cyclone II FPGA. The software portion consists of Verilog code that implements an oscilloscope and function generator in the FPGA. The oscilloscope and generator can be configured to operate in different modes. The project was successful in achieving the desired functionality at acceptable rates of operation, but performance was limited by design decisions and the simplicity of the approach taken.**

**Report Approved by**  
**Project Advisor: \_\_\_\_\_ Date: \_\_\_\_\_**

## Executive Summary

FPGA technology is a growing field in electrical engineering. Classes at Cornell University teach the basics of the system, but lack tools to explore certain applications. This project involved the development of an academic tool to enable the manipulation and investigation of high speed signals.

The goal of the project was to create a bi-directional hardware interface that fed analog values from the outside world into a processor that displayed them on a screen and that created analog values to be transmitted to the outside world. This was realized by a circuit board that housed digital-to-analog and an analog-to-digital conversion chips and an FPGA application that exercised those external interfaces, and displayed the input channel on a VGA monitor.

The hardware design process involved making decisions about how many data to capture and create and how they should be processed when they were in analog form. The software design process involved making decisions about how to store the data and how to perform the digital processing. Small data sizes were chosen to keep the system simple, and data were stored in a cheap static form for display and a more expensive, easily manipulated form during acquisition. Analog processing was kept to a minimum for reliability reasons, and digital processing was performed by logic elements rather than by a fully architected CPU.

The hardware design was divided into two easily separable parts by data direction. The design was largely straightforward, but schematic errors prevented it from being as versatile or robust as it was intended to be. Software design was more complicated because the several systems were required to interact. Each software module handled the operation of part of the system, but the control scheme required that they monitor each other for timing and configuration information.

The outcome of the project was a success. The desired functionality was implemented in full, and could be accurately demonstrated at speeds up to 2MHz. Attempts to achieve higher levels of performance were made difficult by the structure of the design. Future projects will be able to make good use of the results of this one.

## Introduction

Field-Programmable-Gate-Array (FPGA) technology allows engineers a great deal of flexibility in design. The ability to compress large designs into a single chip without investing the time and money required to create a custom integrated circuit greatly simplifies the design process. Even more important is the ability to redesign, add functionality and debug in a matter of hours instead of weeks. FPGAs have gained footing in a wide variety of applications and it important that electrical engineering students gain experience with the technology.

The Terasic DE2 FPGA development kit is used in classes at Cornell University as a platform for learning about basic logic and about FPGA capabilities. The DE2 board is capable of a wide range of functionality, but lacks generic, high-speed digital-to-analog (DAC) and analog-to-digital (ADC) converters. The closest on-board options are the Audio Codec and the VGA output. The codec is limited by a maximum sampling rate of 96kHz, which is far below the FPGA's clock of 50MHz. The VGA output is capable of running at 50MHz but only offers DAC functionality, and is obviously not available if a particular design uses a VGA screen.

To enable the addition of new functionality to the DE2 board, Terasic provides a pair of expansion headers. These two ports each make 36 general-purpose-input/output (GPIO) pins from the FPGA, as well as 3.3V, 5V and Ground power rails available on a standard 100mil header. The FPGA directly controls the GPIO lines, allowing the end user a great deal of flexibility in connecting new components. It is important to note that the Terasic expansion headers do not conform to the Altera Santa Cruz connector form factor and that designs that use the headers will only be usable with DE model boards.

The purpose of this project was to develop DAC and ADC interfaces for the board. These interfaces were created on a daughter card that connects to the expansion headers on the DE2. Coaxial cable sockets connect the interfaces to external signal sources and destinations. An FPGA based oscilloscope application was written to demonstrate the capabilities of the input and output signal paths.

## Design Requirements

The customer for this project is Dr. Bruce Land. He will use the hardware and software developed as an aid in teaching ECE 5760. It offers a starting point for exploring high speed DAC and ADC in a classroom setting. It is considered modifiable even in its final state, as the needs of the classroom dictate. The malleable nature of FPGA firmware dictates that the easiest changes will be new applications for the hardware, but the hardware itself may be adapted to new needs. For these reasons, the software for the oscilloscope shall be written in Verilog for Altera's Quartus II development environment. Cornell owns licenses for the software in the Phillips Hall Digital Lab, and trial versions are available at no cost from Altera. The hardware shall be designed using the gEDA suite of Electronic Design Automation tools. These tools are open source and are available for the Linux operating system free of charge. All hardware and software created in the course of the project shall be delivered to Bruce Land on completion.

Functional project requirements can be divided between the hardware and software components. Hardware shall consist of a daughter card that connects to the DE2 board's expansion headers. It shall host two signal paths, one DAC and one ADC. Each signal path shall offer a selection between a direct path on the analog side, or an amplification stage. The converters shall support positive voltage values only, to simplify the design. The analog side of the daughter card shall feature industry standard BNC terminals to allow the connection of typical oscilloscope probes.

Software shall consist of a digital oscilloscope and function generator implemented on the DE2's Cyclone II FPGA. The oscilloscope shall have user settable parameters, including trigger value, trigger position, and seconds per on-screen division. The oscilloscope output shall be displayed on a VGA computer screen. The screen shall feature indications of the current parameters, as well as static line elements that provide spatial reference on the display.

The function generator shall have three user selectable output modes: sine wave, square wave and triangle wave. The frequency output shall be set by the user between 0 and 16MHz. The

amplitude of the output shall be user selectable as a percentage of the maximum range. The state of the generator shall be displayed on the DE2 board.

Performance requirements were left purposefully vague. Part of the intent of the project was to test what could be done and explore the possible solutions to the problem. With that in mind, the requirements feature the establishment of various performance metrics, rather than dictating minimums that must be met. The ADC and DAC resolution of 8 to 16-bits shall be investigated, and the best option selected. The ADC and DAC shall have a parallel data bus to allow maximum throughput. The ADC and DAC integrated circuits shall support an update rate of at least 50MHz. The oscilloscope and function generator applications shall be exercised across a range of data rates to establish the maximum stable data rates. The VGA screen shall have a resolution of 640x480 pixels.

## **Possible Solutions**

The needs and purpose of the project determined several important aspects of the design. The platform was fixed as the DE2 board. This fixed the required shape and layout of the daughter card and the capabilities of the FPGA that would be used. The capacity of the FPGA was a critical constraint, because that determined how easy it was to fit a particular design into the chip's logic. The Cyclone II is not the most expansive FPGA, and large designs or designs with large, logically unwieldy components may be difficult for the software to fit.

The one of the most important trade offs was the size of the data bus for the ADC and DAC. 1-bit to 24-bits parallel buses are available, but most applications call for a width between 8 and 16-bits. The low end of the scale offers shorter settling times and simplicity, while the high end offers significantly better precision. On the hardware side of the issue, the system was specified as a one sided supply voltage, and priority was placed on finding converters that were specifically intended for such functionality. This implied a simpler unit without a differential analog configuration. On the software side, the oscilloscope must store at least one screen's worth of captured analog values in order

to display them for the VGA. This means that wider data buses will take up more FPGA register space than smaller buses and more logic resources will be required to manipulate them. Since the Cyclone II is not a large FPGA, a smaller bus width would be preferable. With these considerations in mind, several ICs were examined, and an 8 bit, single sided ADC and an equivalent DAC were selected.

The other significant hardware trade off was how to structure the analog signal paths. Adding amplification stages can add flexibility to a design, but also increases the complexity and the number of points of failure in this untested design. A direct connection avoids these pitfalls, but offers no way to manipulate the signals in analog space. Filtering was not considered, as digital filters can be implemented in the FPGA. The solution chosen was to provide both a direct path and an amplified path for each converter. The user can select between them by switching a set of jumpers. This approach combines the benefits of analog manipulation with a backup direct path at little extra design time cost.

The main software design decision was where to put the stored values for display on the FPGA. The Cyclone II FPGA has both internal registers and RAM units as well as a large external RAM. Typical VGA applications use the external ram to store a screen's worth of information, as there is not enough internal memory for a 640x480, 8-bit deep matrix. In this case, however, only one 640-element line of 8-bit values was needed. The stored values correspond not to pixel intensity, but to the magnitude of the voltage at each screen column and which row pixel should be lit. Other on screen elements, such as time and voltage divisions, can be procedurally generated rather than stored. This permits the use of a large array register or of internal memory to store the signal data. Initial tests indicated that the large register constructs were expensive in terms of logic, but allowed fast parallel manipulation and simple programming. Memory, on the other hand, was placed in reserved space on the chip and was easy to synthesize, but could only be read or written once per cycle. Both techniques were eventually used in the final design. A register array was created to read in the ADC values and allow them to be shuffled all at once while the input waited to be triggered. A memory array was

created to allow the VGA controller to use a static copy of the last triggered sequence while a new one was created.

The other major software choice was whether or not to use a NIOS II central processing unit element in the design. The NIOS II is an automatically generated CPU that Altera provides for use with its FPGAs. It allows the user to combine pure C software functionality with the firmware written in Verilog. The programmer can take advantage of the abstraction and sequential execution offered by the C programming language. The disadvantages include the complexity of adding a second programming language and set of source files, and the extra space that the CPU core takes up in the FPGA. It was eventually deemed best to not use the NIOS II. The oscilloscope functionality was not complex enough to warrant the abstraction of the C language, and the default license of the Quartus development environment did not include use of the core. It is possible that a proper license could have been obtained, but time constraints did not permit such delay.

## **Design**

### **Converters**

The design of the daughter card can be broken down by data direction. The DAC is the output and the ADC is the input. The core of each path is an 8-bit converter chip. 8-bits of precision create a quantization range of 0.013V when referenced to 3.3V. As described above, this is sufficient for the purposes of this project. A wide range of integrated circuits supports this precision, offering many possible configurations. The chips that were selected had interfaces that were uncomplicated by optional features and nonessential functionality. They were also selected for their high speed capabilities, achieving at least 50MHz operation. The DAC chip is the HI3338 by Intersil. The ADC chip is the ADC08100 by National Semiconductor. The two chips operate at different voltages (5V for the DAC and 3.3V for the ADC), but both analog segments are referenced based on 3.3V. The analog power pins are fed by a signal that has passed through a low pass filter made up of simple inductor



choke. This helps isolate the analog components from digital noise. Both the analog and digital sides of the system share the same ground. See Appendix B for schematic images and Appendix C for layout images.

## ***DAC***

The DAC is made of two parts, the digital section and the analog section. The digital part consists of the converter chip, its various support capacitors, and its connection off board. The chip is a no-frills digital-to-analog converter that operates at a maximum update rate of 50MHz. It needs bypass capacitors at each of its power pins, but no additional support hardware. The 8 data bits and two control pins are connected to the FPGA via the low-numbered pins of the JP1 header. The control pins are LE, which is a conversion enable pin, and COMP, which activates 2's compliment conversion.

The analog part consists of a set of reference voltages, and the analog line to the output connector. The high reference voltage is set to 3.3V, and the low voltage is set to ground. These values were chosen to keep the DAC's output voltage ranges compatible with the ADC's input voltages and to eliminate the need to generate negative voltages. The analog line splits into two signal paths. The user can select between the two paths by setting the jumpers on CONN1 and CONN2. Connecting pins 3 and 2 selects a path that is terminated for 50Ω, but does not have any active elements. Connecting pins 2 and 1 selects a path that includes an opamp with a gain from 1 to 5. The gain is set via a potentiometer, R4. All of the signal traces for the analog line are routed such that they do not need to change layers or cross traces on other layers. The analog output is a right-angle 50Ω BNC connector. It will allow the user to connect oscilloscope probes directly to the daughter card for testing purposes.

## ***ADC***

The ADC is also made up of two parts, the digital section and the analog section. The digital part consists of the converter chip, its various support hardware, and its connection off board. The chip

is a no-frills analog-to-digital converter that operates at a maximum update rate of 100MHz. It needs bypass capacitors at each of its power pins. The 8 data bits and two control pins are connected to the FPGA via the high-numbered pins of the JP1 header. The control pins are PD, which powers down the chip, and CLK, which synchronizes the conversions with the DE2's main digital clock. The clock signal may require series termination near the pin, which is achieved with a resistor.

The analog part consists of a set of reference voltages, and the analog line to the output connector. The high reference voltage is based on 3.3V, but is divided down to 2V to keep the current through the reference resistor ladder within electrical tolerances. The low voltage is set to ground. These values were chosen to eliminate the need to generate negative voltages. The analog line splits into two signal paths. The user can select between the two paths by setting the jumpers on CONN3 and CONN4. Connecting pins 3 and 2 selects a straight shot path that does not have any active elements. Connecting pins 2 and 1 selects a path that includes an opamp with a gain from 1 to 5 and a voltage divider that can attenuate the signal to 20% of its original value. The gain and attenuation levels are set via pair of potentiometers, R11 and R15 respectively. These variations allow a broad range of inputs to be used without damaging the input. The attenuation, particularly, will allow the opamp gain to be held higher than three. This improves the phase margin in the ADC chip. All of the signal traces for the analog line are routed such that they do not need to change layers or cross traces on other layers. The analog input is a right-angle 50Ω BNC connector. It will allow the user to connect oscilloscope probes directly to the daughter card for testing purposes.

## **Construction**

Constructing the daughter card was not difficult. Solder paste and a toaster oven were used to fix the surface mount components to the board. This method is particularly time saving compared to hand soldering each chip, because the surface tension of the molten solder paste lines the pins up on their pads and ensures a neat connection. The through-hole components were then hand soldered and

the entire board was checked for short circuits. These were also cleaned up by hand. Testing the basic functionality of the board went smoothly. When connected through the direct analog lines, both the DAC and ADC were capable of processing values. More extensive, high-speed tests were done after the software was finished. See Appendix D for photos of the finished board.

Two errors were discovered in this phase of the project. The most obvious problem was that the signal and ground pins of the BNC connectors had been switched. The solution was to flip them to the back side of the board. This had an unfortunate side effect of preventing the structural pins of the connector from mating with the board. The connectors are functional, but delicate. The other error was that the op-amps were not properly configured. They amplify compared to 0Volts, which adds an undesirable DC component to the signal. The DAC op-amp does not work as expected. It produces a strange signal when fed with a sine wave. The ADC side does both amplify and attenuate the signal, but not to the extent hoped. Neither of these problems can be solved with this revision of the board. As the design files will be available to Bruce, future work can rectify these errors.

## **Software**

The software for this project was written in the Verilog hardware description language. The template used was the top level module file and pin assignment file provided by Terasic. These files define a set of signals that control the various components on the DE2 board and connect the signals to the proper FPGA pins. This allowed the use of the names provided by Terasic to connect programmer created logic to the external hardware on the DE2 board.

Several project specific signal sets were created. One set interfaces with the VGA controller. Another controls the function generator. The control logic uses a third set, and the final set holds the state of the data capture logic. One 640 element, 8-bit unpacked array was created to store the ADC data values currently being captured. One reset signal was created and attached to SW[0].

The design consists of 6 main components. The scope control component monitors the input

keys and updates the parameters of the scope as the user changes them. The data capture component records the signal on the ADC data bus according to the parameters of the scope. The VGA controller is a module that handles the timing of the VGA output. The VGA display component selects what to draw to the screen based on the pixel location requested by the VGA controller. The function generator component creates a direct digital synthesis (DDS) output signal based on the setting of the on-board switches and the scope parameters. The parameter display component parses the scope parameters for display on the DE2's seven-segment displays.

### ***Control***

The scope control component allows the user to set the values of 7 scope parameters. They are set using the pushbuttons on the board, called KEYs in the template file. KEYs 0 and 1 cycle through the various settings. KEYs 2 and 3 increment and decrement the currently active setting. The first setting is the number of seconds per vertical division of the screen. There are six selectable values: 1 $\mu$ s, 10 $\mu$ s, 100 $\mu$ s, 1ms, 10ms, and 100ms. The second setting is the trigger value. This value dictates the signal value that must be crossed with a positive slope before the scope will display the signal. It can be set to any integer between 0 and 255. The third setting is the trigger position. This value determines where on the screen the triggered signal will be centered. If it is set to the left edge of the screen, only the data points that are captured after the trigger value will be displayed. As the trigger position moves to the right, more and more data points collected prior to the triggering value will be visible. The trigger position can be set to any x-coordinate on the screen, between 0 and 639. The fourth setting is the trigger mode. It determines whether the scope continually captures new values, or just once. The fifth setting is the source select for the function generator. Three settings are available: sine wave, square wave and triangle wave. When any value is changed, a flag is set to indicate this to the other components. The sixth setting is the frequency of the function generator. It can be changed by 1Hz when SW[16] is off and by 1000Hz when SW[16] is on. The final setting is the amplitude

setting. It determines what fraction of the full 8-bit scale the function generator produces. It can be set between 1 and 1/8 by powers of 1/2. One more control is available. To prevent the scope from capturing new values, toggle SW[17]. It acts as a run-stop switch without changing the parameter settings. See Appendix A for the user's manual.

### ***Capture***

The data capture component collects the values generated by the ADC. It runs on the negative edge of the clock to allow the ADC output enough time to settle before the FPGA samples it. It consists of a state machine with 5 states. The first state initializes the machine so that it is ready to fill an array with captured values. The most important job of the state is to convert the number of seconds per on-screen division into a number of clock cycles between captured values. This ensures that there are always exactly 640 values captured over the period represented on the screen. Based on a screen width of 640 pixels and a sampling rate of 50MHz, the formula for calculating the number of cycles to wait between values is:

$$DelayCount = \frac{50,000,000}{640} \times SecondsPerDivision \times 10$$

*Equation 1: Required Delay Between Samples*

This state immediately transitions to the count state. In the count state, the delay counter increments by one every cycle. If the counter reaches the target, it is reset, and the current value on the ADC output is stored in the capture array. There are two ways the value can be stored in this state. If the array index has not yet reached the trigger position, the value is simply placed in the current index and the index is advanced. The state machine remains in the count state. Once the index reaches the trigger position, a more complicated operation takes place. If the data value is not high enough to trigger the scope, all of the values in the capture array are shifted one element to the left, and the new value is placed immediately behind the trigger position. The state machine remains in the count state. If the trigger

value is met or exceeded by the ADC data, the data are stored in the trigger slot of the array, and the state advances to the triggered state. In the triggered state, the capture array is filled using the same mechanism as in the pre-trigger position portion of the of the count state. When the whole array is full, the state machine transitions back to the copy state to prepare the data for display on the VGA.

The copy state waits for a signal that indicates that the capture array has been copied to memory. That signal is created by a segment of code that runs in sync with the VGA clock instead of the state machine's 50MHz clock. When the VGA driver is in vertical sync, this code increments an index once per cycle. An internal ram module uses the index to update itself from the capture array. When the update is complete, a flag is set so that the capture logic state machine can transition to the delay state. The delay state exists to keep the screen from updating too quickly and to enable the use of the single capture trigger setting. The state machine will pause here for 1/5 of a second, or until the trigger mode is set to auto-run, whichever happens second.

### ***VGA Driver***

The VGA controller is a Verilog module that is provided for the DE2 board by Terasic. It handles the timing and logic for the VGA output. It provides an x-y pair of pixel coordinates to the user and expects the color value of that pixel as an input one cycle later. The value is defined as a 10-bit number for each of red, green, and blue channels of the RGB additive color model. It requires two supplementary modules to operate. One is a phase-locked-loop that converts an on-board 27MHz clock to the 25.17 MHz on which the VGA timings are based. The other provides a delayed reset signal that ensures the VGA controller has enough time to complete its reset. These modules are the only pieces of code that were not written by the author of this report.

### ***VGA Display***

The VGA display component controls what is written to the screen at each pixel. Since the

majority of the screen is blank, and the visible elements can be procedurally generated, only a few pixel values are stored in a VGA memory. This component selects what to draw based on the currently requested pixel from the VGA controller. It can draw 7 elements. The first two are the trigger value and position. The trigger value is drawn as a horizontal red line that the signal must cross to be displayed. The trigger position is drawn as a vertical green line at the x-coordinate of the triggered value. The next three elements make up the displayed signal. The simplest display method for the signal would be to print only the captured values. These values are selected from the VGA memory while the driver is not in a sync state. If the signal transitions faster than one y-pixel per x-pixel, however, there will be vertical gaps between the points. This is solved by connecting adjacent points with vertical lines. These vertical lines make up the other two elements in this group. All three of these elements are printed at maximum white intensity. The final pair of elements is the rule lines. They are faint white vertical and horizontal lines that divide the screen space into 10 equal segments in the x-direction and 8 in the y-direction. They make it easier to read the scope signal. See Appendix E for display photos.

### ***Function Generator***

The function generator is a standard direct digital synthesis device. It is configured to increment a large register each cycle such that the register rolls back to zero with a particular frequency. The frequency is set in Hz by the appropriate scope parameter. The amplitude of the signal is scaled based on the amplitude parameter. Three signals are created by the generator. One is a sine wave. The top 8-bits of the register are indexed into a sine table that generates one sine period each time the register overflows. The other two signals are a square wave and a triangle wave. Both of them switch states when the DDS register reaches either its halfway point or 0. The square wave switches between a 0x00 output and a 0xFF output, and the triangle wave switches between counting up and counting down.

## Parameter Display

The parameter display component is the final component, and is what allows the user to make sense of the scope. For each of the parameters that can be changed, a character representation was created for the DE2's seven-segment display. Two modules were developed to implement this. The `hex_char` module translates a subset ASCII values or decimal digits into appropriate representations in seven segments. One of these was instantiated for each of the 8 HEXx output signals. A `char_gen` module controls the translators. It takes each of the scope parameters as an input and outputs one ASCII or decimal value for each of the `hex_char` instantiations. `char_gen` has a different display for each of the 7 parameters. HEX7 and HEX6 always indicate the setting that is being changed, and the rest of the digits indicate the value. The following table shows the different modes:

Setting	HEX7	HEX6	HEX5	HEX4	HEX3	HEX2	HEX1	HEX0
<b>Second/Division</b>	S	d	[100s]	[10s]	[1s]	E	-	[3 or 6]
<b>Trigger Value</b>	t	r			H (base16)		[16s]	[1s]
<b>Trigger Position</b>	t	P			H (base16)	[256s]	[16s]	[1s]
<b>Trigger State</b>	t	S			[Auto or oncE]			
<b>Source Setting</b>	S	S			[SinE or Squ or tri]			
<b>Hertz setting</b>	H	Z	[1048576s]	[65536s]	[4096s]	[256s]	[16s]	[1s]
<b>Amp Setting</b>	S	A				1	/	[2, 4, 8]

Table 1: Parameter Display Layout

## Testing

The software was more difficult to test than the hardware. Significant reevaluations of how to accomplish tasks took place, as well as considerable debugging. Initially, two huge registers were used to store the capture and display versions of the signal array. The display array was the necessary 640 elements long, but the capture array was twice that size with the trigger value in the center. This allowed the capture logic to be separate from the location of the trigger position. It took approximately 20 minutes for the program to analyze and fit this design into the FPGA. This was prohibitively long for a system that was still in development. The design was optimized down to a single 640 element



capture array with a dynamically determined trigger position, and a memory based display array. This cut compile time down to a more reasonable 10 minutes.

The other major reevaluation was the location of the parameter display. Initially, it was to be part of the VGA display. As the system grew more complex, and the number of parameters increased from 3 to 7, this became difficult to design. A great number of different characters would have to be created, and would require 35 pixels each to be legible. As the display was developed toward the end of the project, this time investment was not feasible. The first alternative considered was the LCD display on the DE2 board. This had the advantage of some internal processing power and pre-made letter graphics, but carried a large overhead in timing and processing requirements. The solution chosen was to use the seven-segment displays on the DE2 board. These suffered from limited display options and a fixed, small number of digits, but offered parallel updates and simple letter graphics.

A problem that surfaced early in the design process was a timing error in the capture logic. The most logically demanding operation that the state machine performed was a simultaneous shift of 639 8-bit values by 1 array position in a single 50MHz clock cycle. Most times when the system triggered, there was no problem. However, when the system updated frequently, erroneous values would appear on the screen often enough to be obvious. After several attempts to revise the logic that was responsible, Quartus was configured to apply maximum effort to fit the design into the FPGA. Since the design did not require maximum resources, the additional timing margin reduced the problem, though it didn't eliminate it.

The most puzzling debug problem had to do with the VGA timing. Seemingly at random, small code changes would cause the VGA display to tear and blur or to recover. It is likely that the complexity of the logic caused timing problems that only appeared under certain fit conditions. Attempts were made to ease the load on the VGA signals, but ultimately only tweaking random code segments seemed to fix the problem.

## Results

The project performed very well overall. The functional requirements were met, with the system able to transmit and receive analog signals at a variety of rates and shapes. It was able to display the signals on a VGA screen with no tearing or timing artifacts. Each of the several settings worked properly, and the parameter display clearly indicated each setting's state.

The project also established performance numbers for the hardware and software. The function generator's DAC output was tested with a real digital oscilloscope to establish maximum rates. Up to 2MHz, the sine output was largely undistorted. It was also fairly accurate, with an error less than 1% across the range of 0 to 2MHz. Above 4MHz, the distortion became much more apparent. Peak to peak measures of the frequency were still very close to the expected values, but the shape was no longer purely sinusoidal. The DDS unit simply progresses too quickly through the sine table, and no longer hits enough values to make a clean sine wave. The DAC shows no sign of flagging, however, and produces reasonable square waves at well over 10MHz.

The scope performs adequately, but suffers significantly from a design decision. At low frequencies it performs very well, accurately capturing sine, square, and triangle waves produced by the DAC. Its selectable parameters allow the user to modify the trigger settings and to change the display time base. A single cycle captured across a 1sec wide screen looks much the same as a cycle captured across a 10 $\mu$ sec wide screen. In between, the signal compresses or expands smoothly as it changes. There is a small source of error in the time bases, as they are determined by the width of the screen and the 50MHz system clock. Time is quantized at 20ns, so the 1 $\mu$ s per division setting is not exact.

This leads into to bigger limitation of the oscilloscope. Since the system was designed to require one data point per screen x-coordinate, a time resolution of greater than 1 $\mu$ s per division is not possible. It would capture less than one point per column. This might not have been insurmountable with extrapolated lines between points, but that would have made the capture or display logic much

more complicated. In addition, such a system would not have produced meaningful output at high frequencies. Theoretically, the system is capable of capturing up to 25MHz before aliasing becomes a problem, but displaying a correct waveform at near half of the sampling rate would require a reconstruction filter, which is beyond the scope of this project.

At a resolution of  $1\mu\text{s}$  per division, the oscilloscope can display a sine wave of up to 2MHz before the peaks and troughs become difficult to distinguish, and a sine wave of up to 4.5MHz before they become impossible to distinguish.

## **Conclusion**

This project forms a solid foundation for further experimentation with high speed ADC and DAC on the DE2 board. Aside from some unfortunate but surmountable design errors, the hardware is an excellent platform for which to develop FPGA applications. It is simple to use and robust against user error. The software design process brought to light many issues to consider when attempting to generate and sample high frequency signals. The software itself gives an example of how to create several systems that interact with and control one another, yet update their status separately. Future work could include attempts to use higher frequency FPGA clocks to coax more performance out of the hardware, or software filtering to improve performance without stressing the hardware.

# Appendix A

## User's Manual

### *Hardware*

The hardware is straightforward to deal with. The first thing to do is remove the plastic shield on the DE2 board. This allows access the expansion headers. Second, with the DE2 turned off, plug the receptacles on the daughter card onto the headers on the DE2 so that the BNC connectors hang out into space. Check to make sure that the receptacles are centered on the headers.

Connect a BNC terminated cable to BNC1 to access the DAC. Connect one to BNC2 to access the ADC. Be careful not to stress the connectors, because the signal and ground pins are the only thing holding them down. Also, be aware that oscilloscope probes often have a switch that toggles attenuation on and off.

To enable the output and input paths, place 100mil jumpers on the four CONN connectors. Placing them toward the outside edge enables the direct paths, which are most reliable. Placing them toward the inside of the board enables the amplified paths, which are not fully operational. To adjust the DAC output gain, turn the potentiometer labeled R4. To adjust the attenuation factor on the ADC, turn R15. To adjust the input gain for on the ADC, turn R11.

### *Software*

The controls of the oscilloscope and function generator are the four KEYS, SW[17], SW[16] and SW[1]. SW[1] is a toggleable reset. It brings the scope settings to their default values, but doesn't clear the last captured scope trace. SW[17] is Run-Stop. When it is on, the scope retains its settings, but does not update the screen.

To change the system settings, use the four KEYS. KEY[0] and KEY[1] change which parameter is being edited. KEY[2] and KEY[3] decrement and increment the parameter respectively. When the DDS frequency is being set, SW[16] can be turned on to change the increment to 1000Hz. See Table 1 for the details of the parameter display.

# Appendix B

## Schematic Images

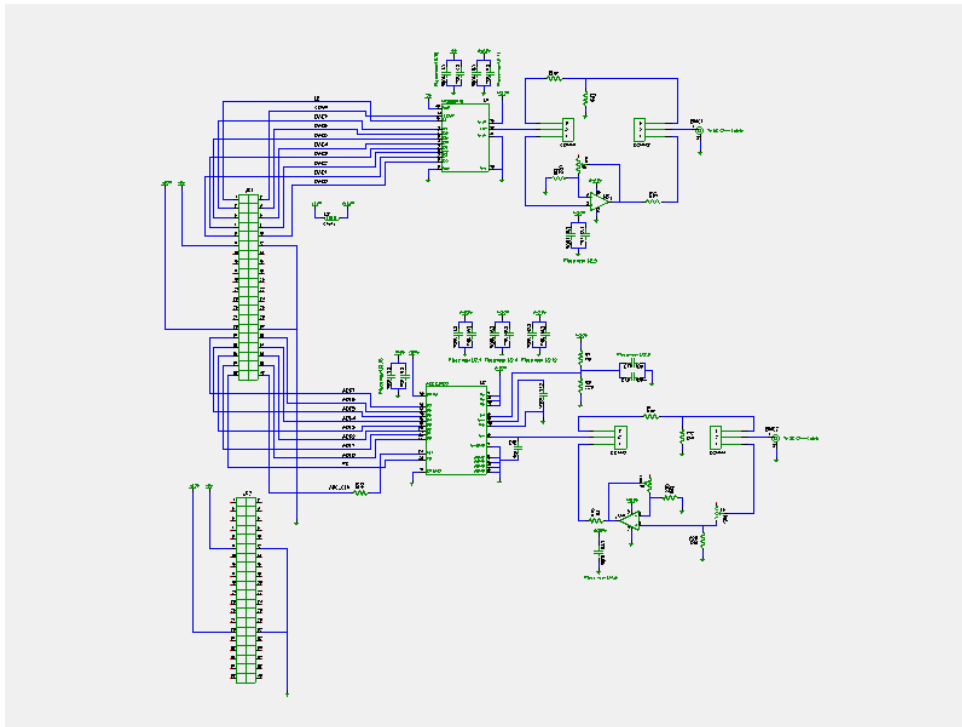


Figure B1: Schematic Overview

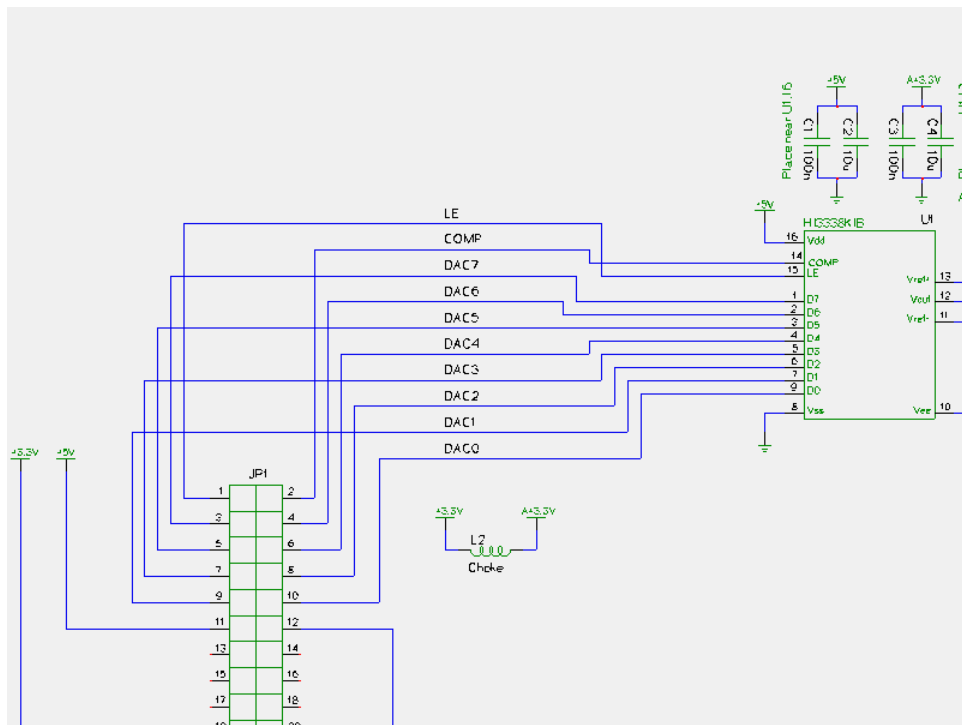


Figure B2: DAC Digital Side

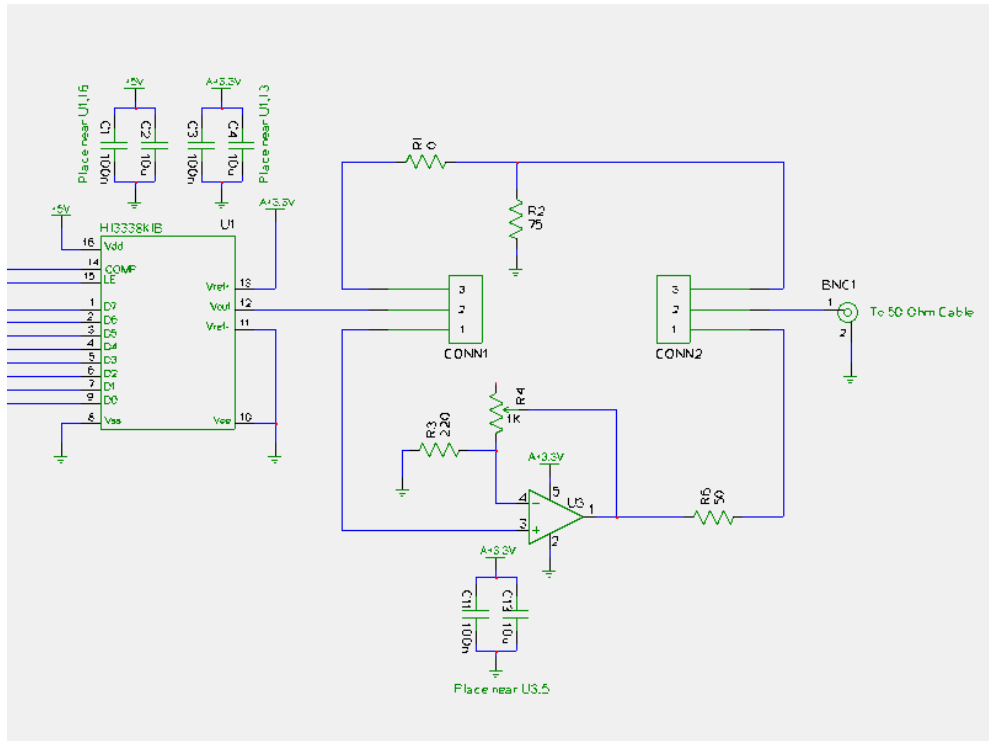


Figure B3: DAC Analog Side

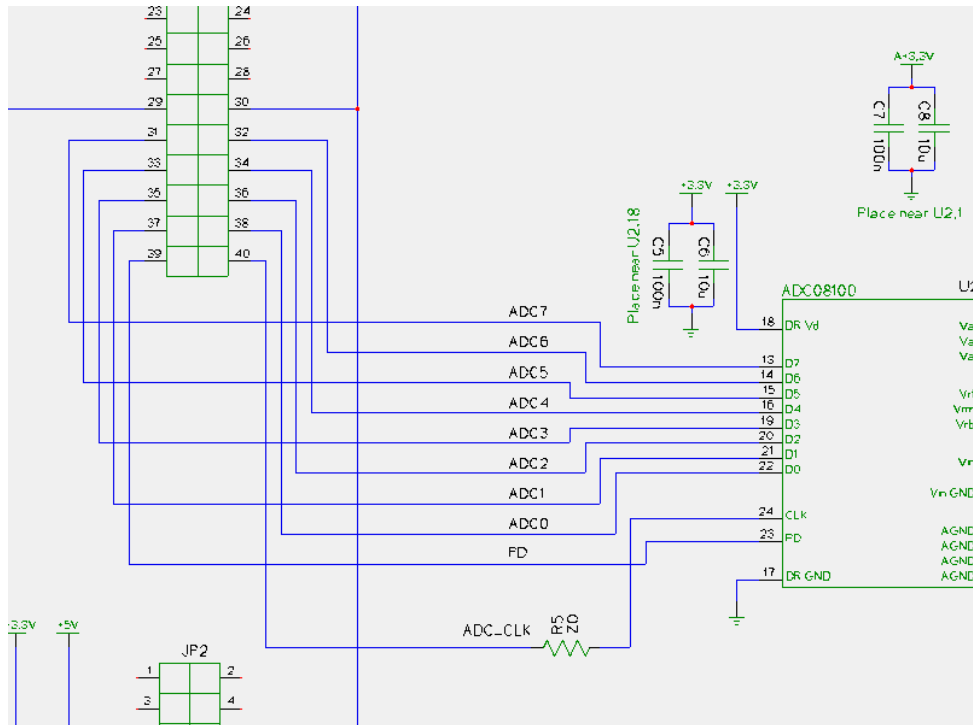


Figure B4: ADC Digital Side

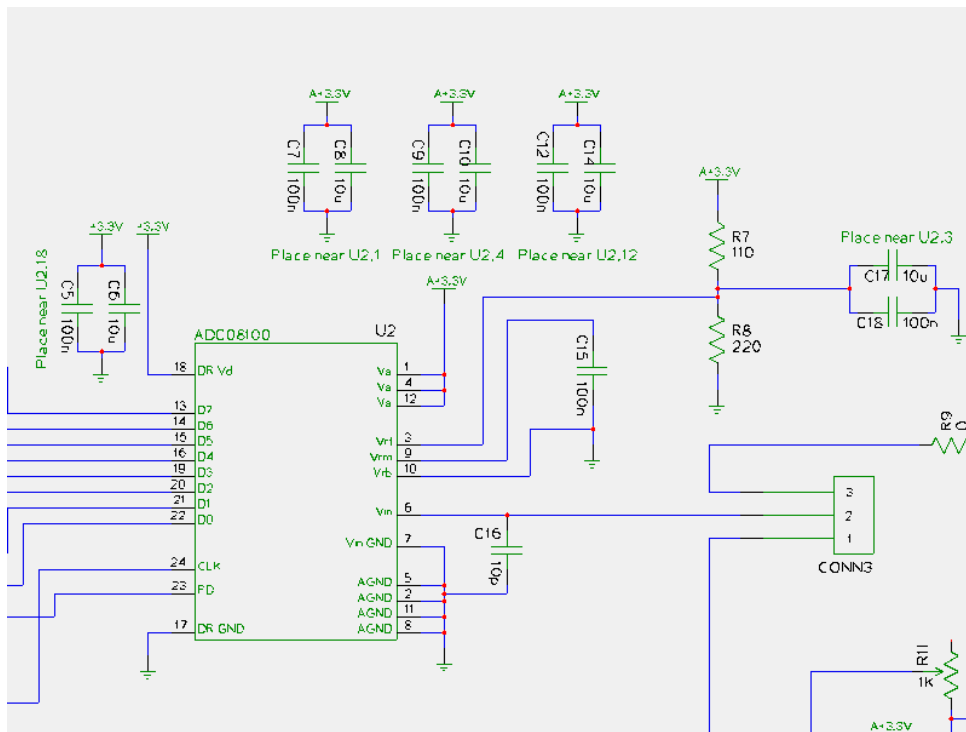


Figure B5: ADC Detail

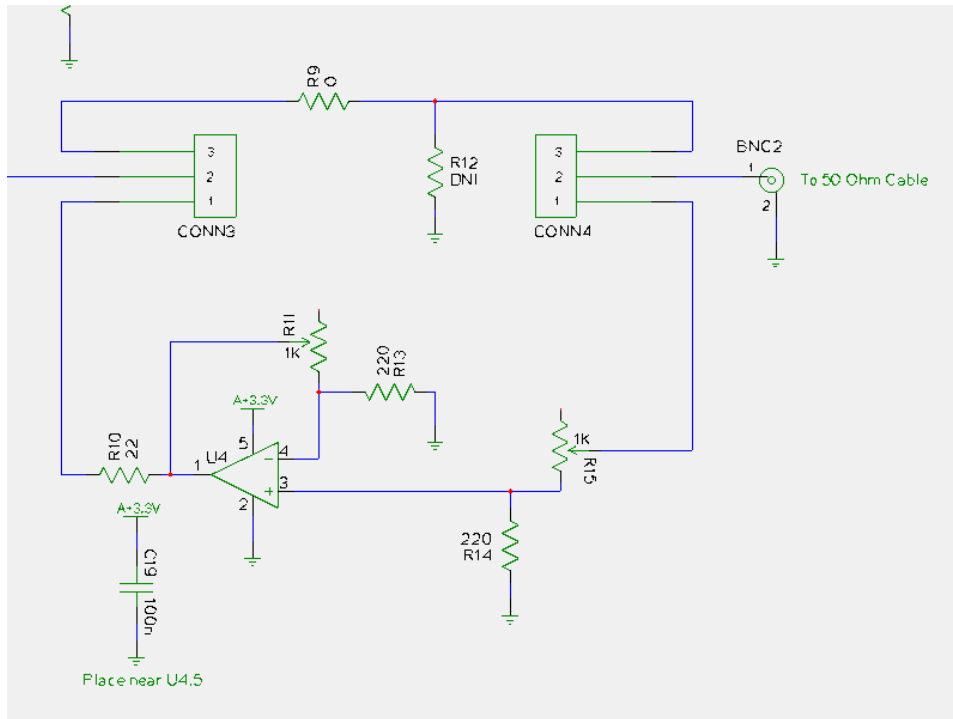


Figure B6: ADC Analog Side

# Appendix C

## Layout Images

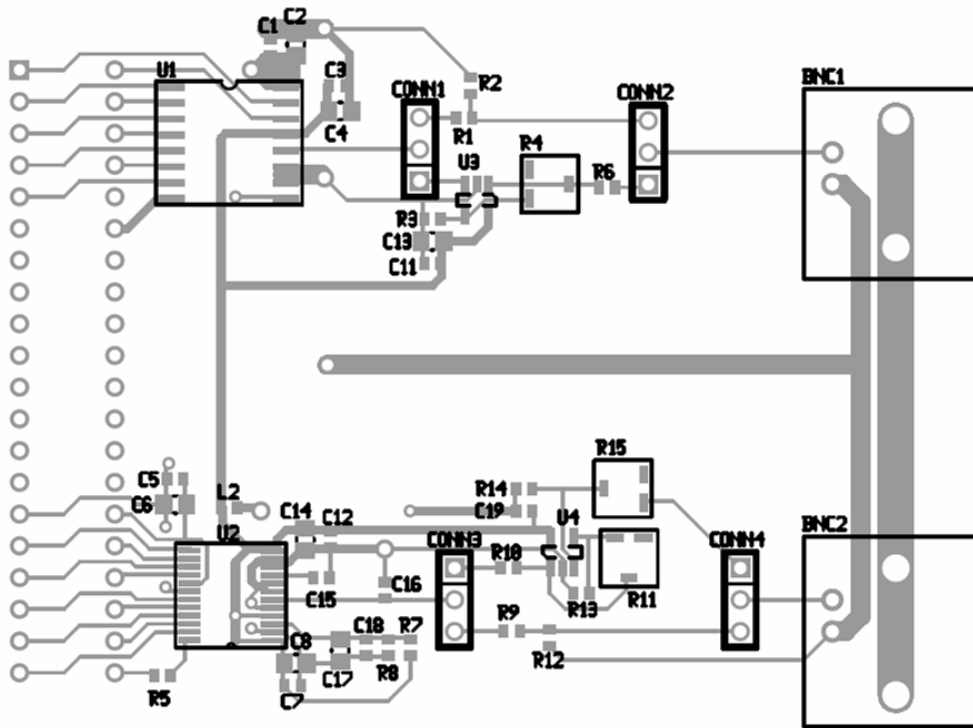


Figure C1: Top Layout

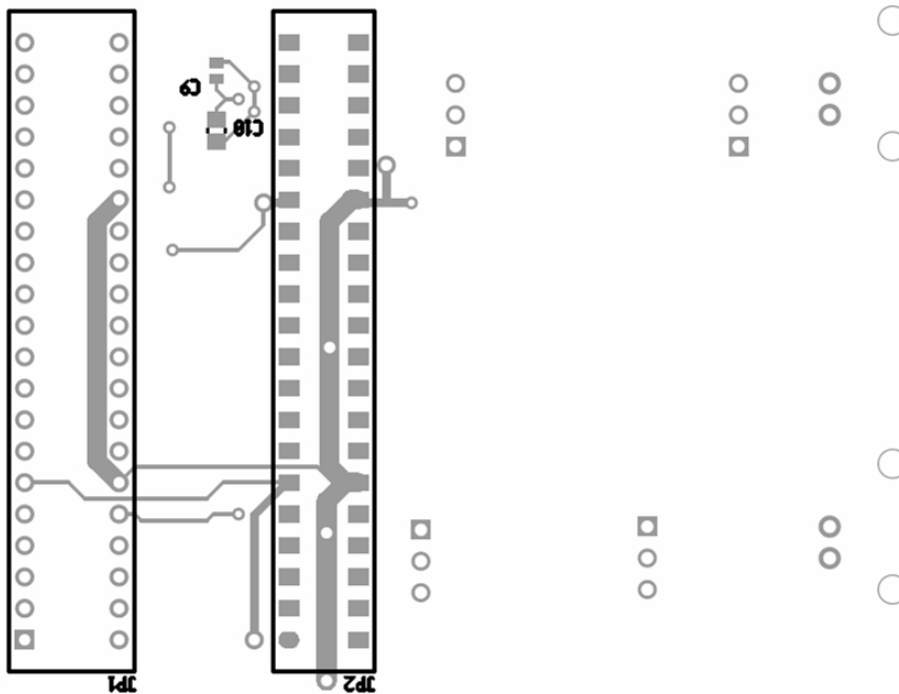
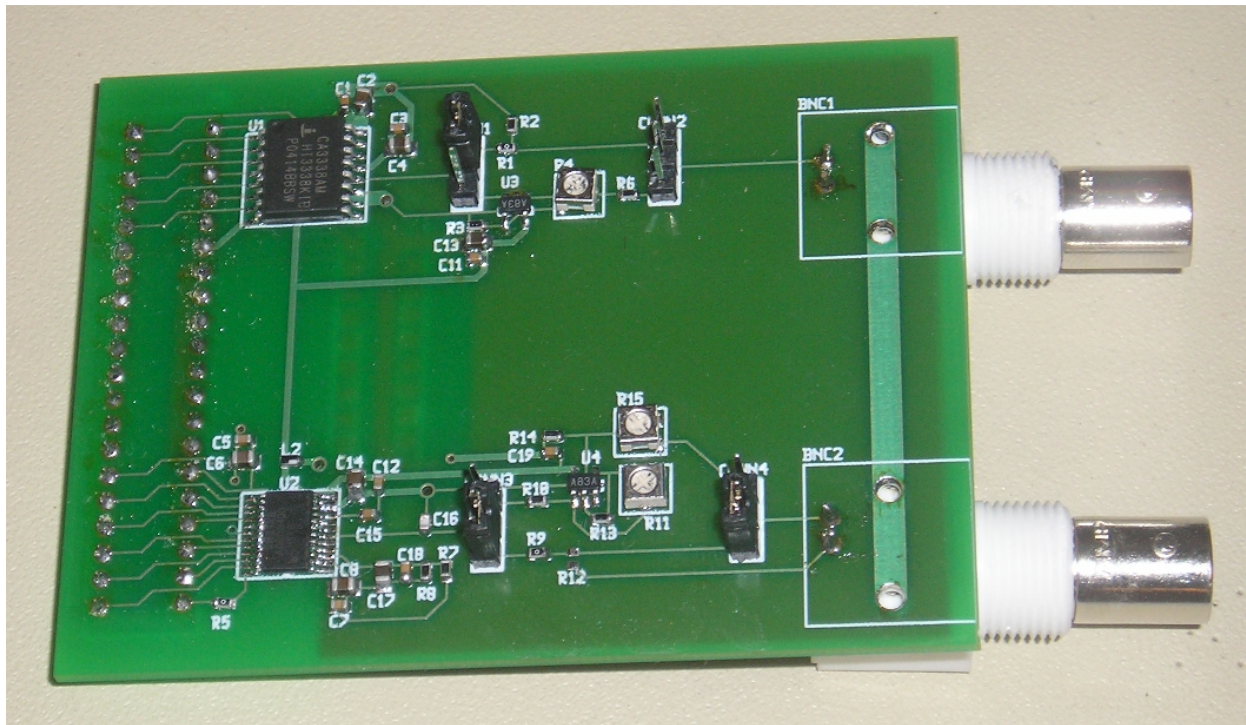


Figure C2: Bottom Layout

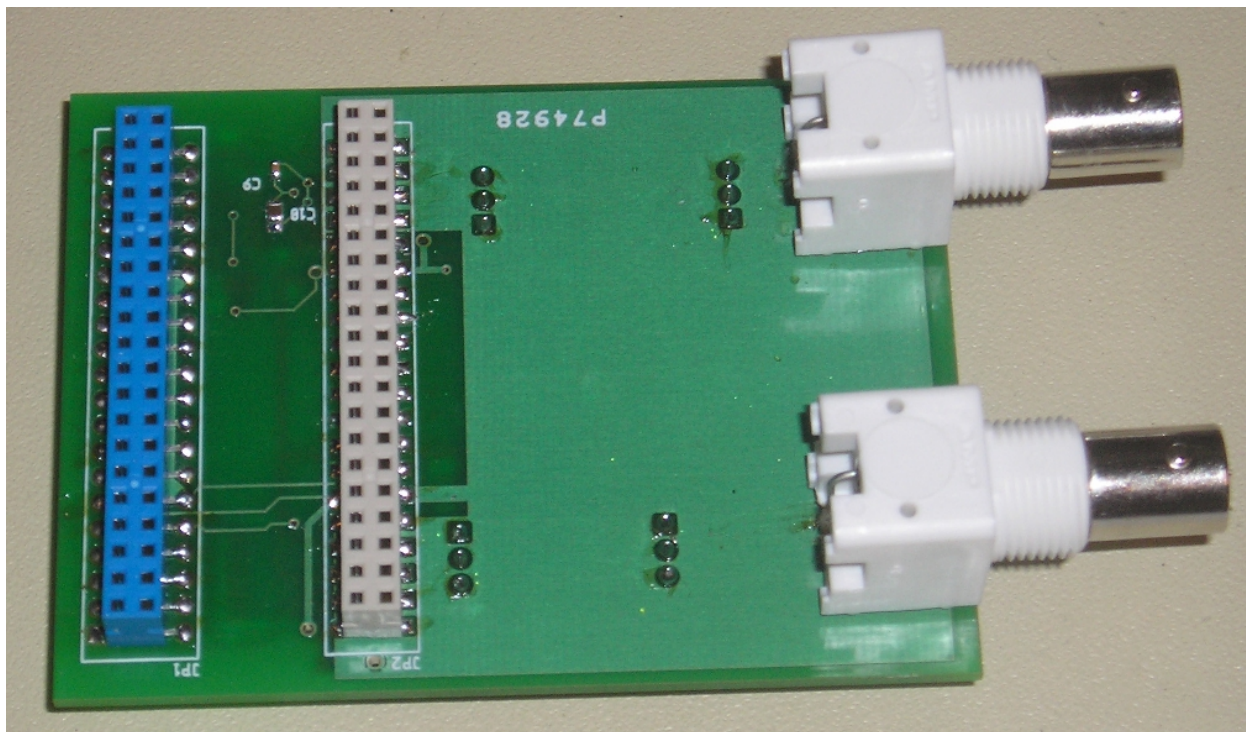


## Appendix D

### Board Photos



*Figure D1: Top Side of the Assembled Board*



*Figure D2: Bottom Side of the Assembled Board*

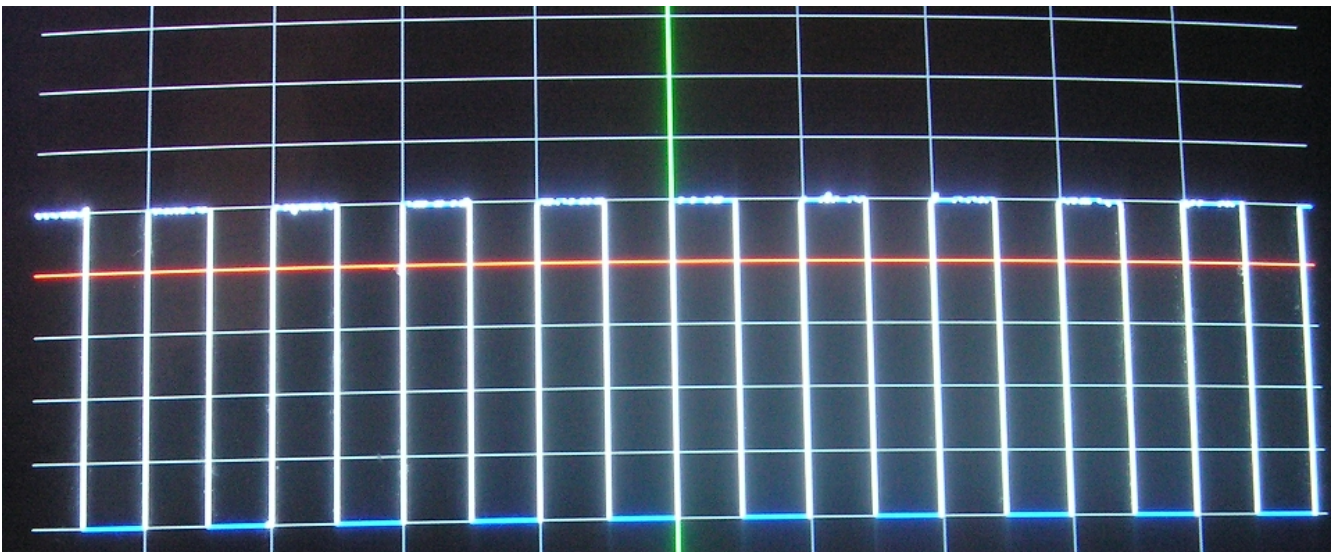


## Appendix E

### Display Photos



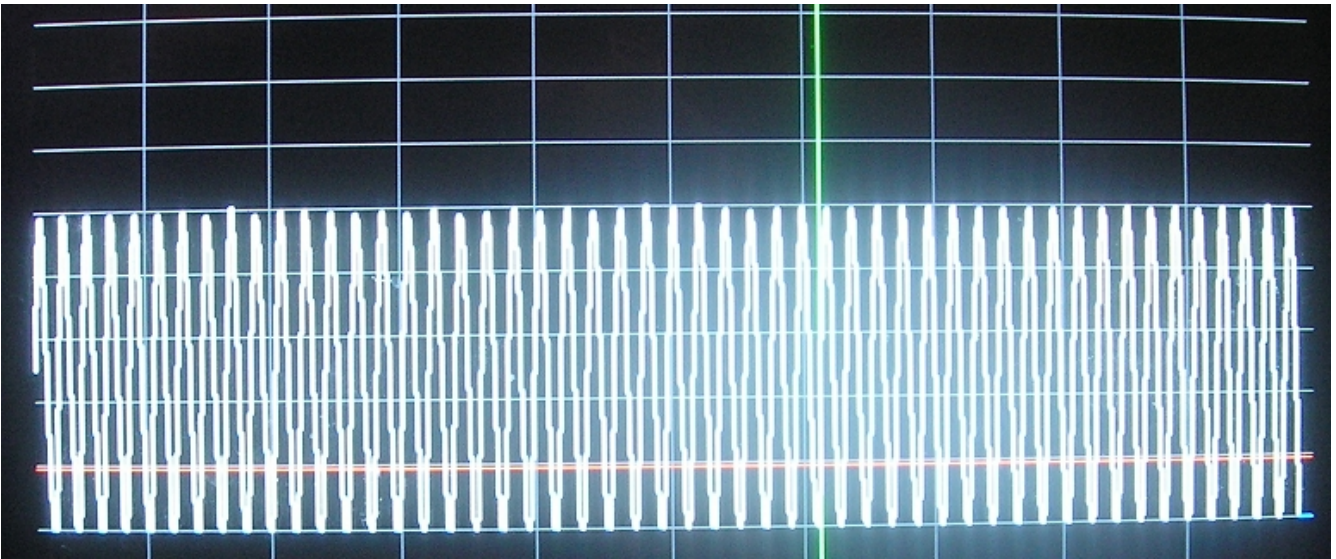
*Figure E1: 10Hz Sine Wave at 100ms per Division*



*Figure E2: 10Hz Square Wave at 100ms per Division*



*Figure E3: 10Hz Triangle Wave at 100ms per Division*



*Figure E4: 2MHz Sine Wave at 1 $\mu$ s per Division*

## Appendix F

### Code

#### *DE2\_TOP.v (Header removed for copyright reasons)*

```
//reordered aliases for the DAC and ADC GPIO pins
wire [7:0] fastADC;
wire [7:0] fastDAC;
reg [7:0] oldFastADC;

//Scope control signals
reg [3:0] settingSelect;
reg [24:0] controlDelay;
reg [2:0] secPerDiv;
reg [3:0] DDSsource;
reg [23:0] hzVal;
reg [3:0] ampVal;
reg [7:0] triggerVal;
reg [9:0] triggerPos;
reg [3:0] triggerSetting;
reg newSetting;
reg [3:0] stateControl;
wire runStop;

//VGA Elements
wire DELAYED_RST;
wire VGA_CTRL_CLK;
wire AUD_CTRL_CLK;
wire [9:0] mVGA_R;
wire [9:0] mVGA_G;
wire [9:0] mVGA_B;
wire [19:0] mVGA_ADDR; //video memory address
wire [9:0] Coord_X, Coord_Y; //display coords
reg [3:0] stateVGA;
reg [9:0] signalRed;
reg [9:0] signalGreen;
reg [9:0] signalBlue;
reg memTransferDone;

//Scope capture signals
reg [16:0] captureCounter;
reg [16:0] counterTarget;
reg [10:0] arrayPosition;
reg [3:0] stateScope;
reg [9:0] i;
reg [9:0] j;
reg [25:0] updateDelay;

//Display elements
wire [7:0] hexDisp [7:0];

//Signal generator elements
reg [7:0] sine;
wire [7:0] square;
wire [7:0] triangle;
wire [7:0] sinAddr;
reg [31:0] sinDDS;
wire [31:0] ddsInc;
wire [7:0] DDSOut;

//Array to hold captured values
reg [7:0] capturedVals [0:639];
wire [7:0] valForVGA;
reg [7:0] oldValForVGA;

//Scope states
parameter init=4'd1, count=4'd2, triggered=4'd3, copy=4'd4, delay=5'd5;

//Control states
parameter idleControl=4'd0, delayControl=4'd1;
```



```

//Setting selected
parameter setSecPerDiv=4'd0, setTriggerVal=4'd1, setTriggerPos=4'd2, setTriggerSetting=4'd3,
setSource=4'd4, setHz=4'd5, setAmpVal=4'd6;

//Function generated
parameter makeSine=4'd0, makeSquare=4'd1, makeTriangle=4'd2;

//Triggering mode
parameter freeRunning=4'd0, once=4'd1;

// Turn on all display
assign LEDG      = {1'h1, fastDAC};
assign LEDR      = sinDDS[31:14];
assign LCD_ON    = 1'b1;
assign LCD_BLON  = 1'b1;

// All inout port turn to tri-state
assign DRAM_DQ   = 16'hzzzz;
assign FL_DQ     = 8'hzz;
assign SRAM_DQ   = 16'hzzzz;
assign OTG_DATA  = 16'hzzzz;
assign LCD_DATA  = 8'hzz;
assign SD_DAT    = 1'bz;
assign I2C_SDAT  = 1'bz;
assign ENET_DATA = 16'hzzzz;
assign AUD_ADCLRCK = 1'bz;
assign AUD_DACLCK = 1'bz;
assign AUD_BCLK  = 1'bz;
assign GPIO_1    = 36'hzzzzzzzz;
assign GPIO_0    = 36'hzzzzzzzz;

assign TD_RESET  = 1'b1; //Enable 27 MHz

assign runStop   = (SW[17]);

//set up the GPIO pins to run the daughter card
assign GPIO_0[35:34] = {CLOCK_50, 1'b0};
assign fastADC       = {GPIO_0[26], GPIO_0[27], GPIO_0[28], GPIO_0[29],
                       GPIO_0[30], GPIO_0[31], GPIO_0[32], GPIO_0[33]};
assign GPIO_0[25:10] = 16'hzzzz;
assign GPIO_0[9:2]   = {fastDAC[0], fastDAC[1], fastDAC[2], fastDAC[3],
                       fastDAC[4], fastDAC[5], fastDAC[6], fastDAC[7]};
assign GPIO_0[1:0]   = 2'b00;

//Output is the DDS output scaled by Amplitude value
assign fastDAC       = DDSOut >> ampVal;

//VGA signals
assign mVGA_R = signalRed;
assign mVGA_G = signalGreen;
assign mVGA_B = signalBlue;

//Signal Gennerator DDS increment
assign ddsInc = (hzVal * 159); //159= 2^32/27000000 so hzVal is in Hz

//For VGA reset
Reset_Delay r0 ( .iCLK(CLOCK_50), .oRESET(DELAYED_RST));

//For VGA CLK
VGA_Audio_PLL p1 ( .areset(~DELAYED_RST), .inclk0(CLOCK_27), .c0(VGA_CTRL_CLK), .c1(AUD_CTRL_CLK), .c2(VGA_CLK));

//For Driving VGA
VGA_Controller u1 ( // Host Side
                  .iCursor_RGB_EN(4'b0111),
                  .oAddress(mVGA_ADDR),
                  .oCoord_X(Coord_X),
                  .oCoord_Y(Coord_Y),
                  .iRed(mVGA_R),
                  .iGreen(mVGA_G),
                  .iBlue(mVGA_B),
                  // VGA Side
                  .oVGA_R(VGA_R),
                  .oVGA_G(VGA_G),

```

```

        .oVGA_B(VGA_B),
        .oVGA_H_SYNC(VGA_HS),
        .oVGA_V_SYNC(VGA_VS),
        .oVGA_SYNC(VGA_SYNC),
        .oVGA_BLANK(VGA_BLANK),
        // Control Signal
        .iCLK(VGA_CTRL_CLK),
        .iRST_N(DELAYED_RST));

//Ram to store signal vals for VGA
//The logic to run this is imbedded in the instantiation. Nice, huh?
//Output always goes to the display logic
//Input slided along the capture array
//Address is the capture array index or the X coord
//Write enable is t when we're in the Vertical sync and the scope is in the copy state
ram_unit r1
(
    .q(valForVGA),
    .d(capturedVals[j]),
    .address((~VGA_VS || ~VGA_HS) ? j : Coord_X),
    .we((stateScope == copy && ~VGA_VS) ? 1 : 0),
    .clk(VGA_CTRL_CLK)
);

//Character generator for parameter display
char_gen c1
(
    .hexDisp0(hexDisp[0]),
    .hexDisp1(hexDisp[1]),
    .hexDisp2(hexDisp[2]),
    .hexDisp3(hexDisp[3]),
    .hexDisp4(hexDisp[4]),
    .hexDisp5(hexDisp[5]),
    .hexDisp6(hexDisp[6]),
    .hexDisp7(hexDisp[7]),
    .hzVal(hzVal),
    .ampVal(ampVal),
    .triggerVal(triggerVal),
    .triggerPos(triggerPos),
    .triggerSetting(triggerSetting),
    .DDSSource(DDSSource),
    .secPerDiv(secPerDiv),
    .settingSelect(settingSelect)
);

//Convert ASCII to HEX output
hex_char h0
(
    .hexDisp(HEX0),
    .value(hexDisp[0]),
);

hex_char h1
(
    .hexDisp(HEX1),
    .value(hexDisp[1]),
);

hex_char h2
(
    .hexDisp(HEX2),
    .value(hexDisp[2]),
);

hex_char h3
(
    .hexDisp(HEX3),
    .value(hexDisp[3]),
);

hex_char h4
(
    .hexDisp(HEX4),
    .value(hexDisp[4]),
);

hex_char h5
(
    .hexDisp(HEX5),
    .value(hexDisp[5]),
);

```

```

hex_char h6          (
                    .hexDisp (HEX6),
                    .value (hexDisp[6]),
                    );

hex_char h7          (
                    .hexDisp (HEX7),
                    .value (hexDisp[7]),
                    );

//Scope control
always @ (posedge CLOCK_50)
begin
    if (~DELAYED_RST || SW[0])
    begin
        triggerVal      <= 8'd128;
        triggerPos      <= 10'd320;
        secPerDiv       <= 3'b0;
        controlDelay    <= 25'd0;
        settingSelect   <= 4'b0;
        newSetting      <= 1'b0;
        hzVal           <= 24'h0;
        ampVal          <= 4'h0;
        triggerSetting  <= 4'b0;
        stateControl    <= idleControl;
        settingSelect   <= setSecPerDiv;
        DDSSource      <= 4'b0;
    end
    //no input when stopped
    else if (runStop)
    begin
        stateControl    <= idleControl;
    end
    else
    begin
        //control state machine
        case(stateControl)
            idleControl:
            begin
                //reset delay value
                controlDelay <= 25'd0;
                //if a key 0 or 1 is pressed, change which setting is selected
                if (KEY[0] == 0 && settingSelect < 6)
                begin
                    settingSelect <= settingSelect + 1;
                    newSetting    <= 1;
                    stateControl  <= delayControl;
                end
                else if (KEY[1] == 0 && settingSelect > 0)
                begin
                    settingSelect <= settingSelect - 1;
                    newSetting    <= 1;
                    stateControl  <= delayControl;
                end
                else
                begin
                    //If key 2 or 3 is pressed, increment or decrement the selected
                    parameter
                    case(settingSelect)
                        //seconds per division
                        setSecPerDiv:
                        begin
                            if (KEY[2] == 0 && secPerDiv < 5)
                            begin
                                secPerDiv <= secPerDiv + 1;
                                newSetting <= 1;
                                stateControl <= delayControl;
                            end
                            else if (KEY[3] == 0 && secPerDiv > 0)
                            begin
                                secPerDiv <= secPerDiv - 1;
                                newSetting <= 1;
                                stateControl <= delayControl;
                            end
                        end
                    end
                end
            end
        end
    end
end

```

parameters

```
//trigger value
setTriggerVal:
begin
    if (KEY[2] == 0 && triggerVal < 255)
    begin
        triggerVal <= triggerVal + 1;
        newSetting <= 1;
        controlDelay <= 25'd28000000;
        stateControl <= delayControl;
    end
    else if (KEY[3] == 0 && triggerVal > 0)
    begin
        triggerVal <= triggerVal - 1;
        newSetting <= 1;
        controlDelay <= 25'd28000000;
        stateControl <= delayControl;
    end
end

//trigger position
setTriggerPos:
begin
    if (KEY[2] == 0 && triggerPos < 639)
    begin
        triggerPos <= triggerPos + 1;
        newSetting <= 1;
        //Don't wait as long between changes of some
        controlDelay <= 25'd28000000;
        stateControl <= delayControl;
    end
    else if (KEY[3] == 0 && triggerPos > 0)
    begin
        triggerPos <= triggerPos - 1;
        newSetting <= 1;
        controlDelay <= 25'd28000000;
        stateControl <= delayControl;
    end
end

//Trigger mode
setTriggerSetting:
begin
    if (KEY[2] == 0 && triggerSetting < 1)
    begin
        triggerSetting <= triggerSetting + 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
    else if (KEY[3] == 0 && triggerSetting > 0)
    begin
        triggerSetting <= triggerSetting - 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
end

//DDS source
setSource:
begin
    if (KEY[2] == 0 && DDSSource < 2)
    begin
        DDSSource <= DDSSource + 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
    else if (KEY[3] == 0 && DDSSource > 0)
    begin
        DDSSource <= DDSSource - 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
end
```



```

//frequency setting
setHz:
begin
    (hzVal < 15999001 && SW[16]))
        if (KEY[2] == 0 && ((hzVal < 16000000 && ~SW[16]) ||
            begin
                //Change frequency faster when SW 16 is on
                if (SW[16])
                begin
                    hzVal <= hzVal + 1000;
                    controlDelay <= 25'd29500000;
                end
                else
                begin
                    hzVal <= hzVal + 1;
                    controlDelay <= 25'd29100000;
                end
                newSetting <= 1;
                stateControl <= delayControl;
            end
            else if (KEY[3] == 0 && ((hzVal > 0 && ~SW[16]) ||
                begin
                    if (SW[16])
                    begin
                        hzVal <= hzVal - 1000;
                        controlDelay <= 25'd29500000;
                    end
                    else
                    begin
                        hzVal <= hzVal - 1;
                        controlDelay <= 25'd29100000;
                    end
                    newSetting <= 1;
                    stateControl <= delayControl;
                end
            end
        end

//Set amplitude divider
setAmpVal:
begin
    if (KEY[2] == 0 && ampVal < 3)
    begin
        ampVal <= ampVal + 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
    else if (KEY[3] == 0 && ampVal > 0)
    begin
        ampVal <= ampVal - 1;
        newSetting <= 1;
        stateControl <= delayControl;
    end
end

default: settingSelect <= setSecPerDiv;
endcase
end

//Wait here to keep settings from being changed too fast
delayControl:
begin
    newSetting <= 0;
    if (controlDelay <= 25'd30000000)
    begin
        controlDelay <= controlDelay + 1;
        stateControl <= delayControl;
    end
    else
    begin
        stateControl <= idleControl;
    end
end
end

```

```

                default: stateControl <= idleControl;
            endcase
        end
    end

//VGA display
always @ (posedge VGA_CTRL_CLK)
begin
    oldValForVGA <= valForVGA;
    //display the points that are stored
    if ((Coord_Y == (360 - valForVGA)))
    begin
        signalRed <= 10'h000;
        signalBlue <= 10'h3FF;
        signalGreen <= 10'h000;
    end
    //connect adjacent points with vertical lines
    else if ((360 - oldValForVGA) <= Coord_Y) && (Coord_Y <= (360 - valForVGA))
    begin
        signalRed <= 10'h3FF;
        signalBlue <= 10'h3FF;
        signalGreen <= 10'h3FF;
    end
    //connect adjacent points with vertical lines
    else if ((360 - oldValForVGA) >= Coord_Y) && (Coord_Y >= (360 - valForVGA))
    begin
        signalRed <= 10'h3FF;
        signalBlue <= 10'h3FF;
        signalGreen <= 10'h3FF;
    end
    //display the trigger value line
    else if (Coord_Y == (360 - triggerVal))
    begin
        signalRed <= 10'h3FF;
        signalBlue <= 10'h000;
        signalGreen <= 10'h000;
    end
    //display the trigger position line
    else if (Coord_X == triggerPos)
    begin
        signalRed <= 10'h000;
        signalBlue <= 10'h000;
        signalGreen <= 10'h3FF;
    end
    //display the vertical guide lines
    else if ((Coord_X == 0) || (Coord_X == 64) || (Coord_X == 128) || (Coord_X == 192) || (Coord_X
== 256) || (Coord_X == 320) || (Coord_X == 384) || (Coord_X == 448) || (Coord_X == 512) || (Coord_X ==
576))
    begin
        signalRed <= 10'h0FF;
        signalBlue <= 10'h0FF;
        signalGreen <= 10'h0FF;
    end
    //display the horizontal guide lines
    else if ((Coord_Y == 360) || (Coord_Y == 328) || (Coord_Y == 296) || (Coord_Y == 264) ||
(Coord_Y == 232) || (Coord_Y == 200) || (Coord_Y == 168) || (Coord_Y == 136) || (Coord_Y == 104))
    begin
        signalRed <= 10'h0FF;
        signalBlue <= 10'h0FF;
        signalGreen <= 10'h0FF;
    end
    //Everything else is black
    else
    begin
        signalRed <= 10'h000;
        signalBlue <= 10'h000;
        signalGreen <= 10'h000;
    end
end

//Capture data values
always @ (negedge CLOCK_50)
begin
    if (runStop | newSetting)

```

```

begin
    captureCounter    <= 17'b0;
    arrayPosition     <= 11'b0;
    counterTarget     <= 17'b0;
    oldFastADC        <= 8'b0;
    updateDelay       <= 26'b0;
    stateScope        <= init;
end

else
begin
    case(stateScope)
        //Set up the time between samples
        init:
        begin
            arrayPosition    <= 11'b0;
            captureCounter <= 0;
            case(secPerDiv)
                3'd0:
                begin
                    counterTarget <= 17'd78125;
                end

                3'd1:
                begin
                    counterTarget <= 17'd7813;
                end

                3'd2:
                begin
                    counterTarget <= 17'd781;
                end

                3'd3:
                begin
                    counterTarget <= 17'd78;
                end

                3'd4:
                begin
                    counterTarget <= 17'd7;
                end

                3'd5:
                begin
                    counterTarget <= 17'd1;
                end

                default: counterTarget <= 17'd78125;
            endcase
            stateScope <= count;
        end

        //Wait until time between samples has elapsed
        count:
        begin
            captureCounter <= captureCounter + 1;
            if (captureCounter == counterTarget)
            begin
                captureCounter <= 0;
                //if we havent gotten to the trigger position yet
                if (arrayPosition <= (triggerPos - 1))
                begin
                    //capture the value and go to the next position
                    capturedVals[arrayPosition] <= fastADC;
                    arrayPosition <= arrayPosition + 1;
                    oldFastADC <= fastADC;
                    stateScope <= count;
                end
                //if we're at the trigger position and the value crossed it
                else if ((fastADC >= triggerVal) && (oldFastADC < triggerVal))
                begin
                    //capture the value and go to the triggered state
                    capturedVals[arrayPosition] <= fastADC;
                end
            end
        end
    endcase
end

```

```

        oldFastADC <= fastADC;
        stateScope <= triggered;
    end
    else
    begin
        //if the value is not trigger worthy, shift everything down
        //and then store the value
        //This is probably the ugliest code bit here
        for (i=1; i<=639; i=i+1)
        begin
            capturedVals[i-1] <= capturedVals[i];
        end
        capturedVals[(triggerPos - 1)] <= fastADC;
        oldFastADC <= fastADC;
        stateScope <= count;
    end
end
end

//We've gotten a good trigger value
triggered:
begin
    captureCounter <= captureCounter + 1;
    if (captureCounter == counterTarget)
    begin
        captureCounter <= 0;
        //Fill out the rest of the array with whatever comes in
        //then go to the copy state
        if (arrayPosition <= 639)
        begin
            capturedVals[arrayPosition] <= fastADC;
            arrayPosition <= arrayPosition + 1;
            stateScope <= triggered;
        end
        else
        begin
            stateScope <= copy;
        end
    end
end
end

//Wait until memory has been updated
copy:
begin
    if (memTransferDone)
    begin
        updateDelay = 0;
        stateScope <= delay;
    end
    else
    begin
        stateScope <= copy;
    end
end
end

//wait for 1/20 of a second to keep the update rate from getting too fast
//also, if we're in the "once" trigger mode, wait until we get out of it
//then go back to init
delay:
begin
    if ((updateDelay <= 10000000) || (triggerSetting == once))
    begin
        stateScope <= delay;
        updateDelay = updateDelay+1;
    end
    else
        stateScope <= init;
    end
end

default: stateScope <= init;

endcase
end

```

```

end

//VGA memory update
//needed so the memory ram can run at VGA clock rate
always @ (posedge VGA_CTRL_CLK)
begin
    if (runStop | newSetting)
        begin
            memTransferDone <= 0;
            j <= 0;
        end
    else if (memTransferDone)
        begin
            j <= 0;
            memTransferDone <= 0;
        end
    end
    //if we're in the scope copy state and it's a Veritcal sync
    else if ((stateScope == copy) && (~VGA_VS))
        begin
            if ( j <= 10'd639 )
                begin
                    //increment through the screen's x indices
                    j <= j+1;
                end
            else
                begin
                    //When we're done, let the copy state know
                    memTransferDone <= 1;
                end
            end
        end
    end
end

//DDS updatar on CLOCK_27 b/c 50MHz sampling rate
always@(negedge CLOCK_27)
begin
    sinDDS <=      sinDDS + ddsInc ;
end

//Sine wave generator
assign sinAddr = sinDDS[31:24];

//Square wave generator
assign square = ((sinDDS[31] == 1) ? 8'hFF : 8'h00);

//Triangle wave generator
assign triangle = ((sinDDS[31] == 1) ?  sinDDS[31:24] : 8'hFF - sinDDS[31:24]);

assign DDSOut = (DDSSource == makeSine)      ?  sine:
                ((DDSSource == makeTriangle) ?  triangle:
                ((DDSSource == makeSquare)    ?  square: 0));

//Sin Wave ROM Table
always@(sinAddr)
begin
    case(sinAddr)
        8'h00: sine = 8'hFF ;
        8'h01: sine = 8'hFE ;
        8'h02: sine = 8'hFE ;
        8'h03: sine = 8'hFE ;
        .//Values removed to save space
        .
        .
        8'hfe: sine = 8'hFE ;
        8'hff: sine = 8'hFE ;
        default: sine = 8'h0;
    endcase
end

endmodule

```

## char\_gen.v

```
module char_gen(  
  
    output reg [7:0] hexDisp0,  
    output reg [7:0] hexDisp1,  
    output reg [7:0] hexDisp2,  
    output reg [7:0] hexDisp3,  
    output reg [7:0] hexDisp4,  
    output reg [7:0] hexDisp5,  
    output reg [7:0] hexDisp6,  
    output reg [7:0] hexDisp7,  
    input [23:0] hzVal,  
    input [3:0] ampVal,  
    input [7:0] triggerVal,  
    input [9:0] triggerPos,  
    input [3:0] triggerSetting,  
    input [3:0] DDSSource,  
    input [3:0] secPerDiv,  
    input [3:0] settingSelect  
);  
  
//Setting selected  
parameter setSecPerDiv=4'd0, setTriggerVal=4'd1, setTriggerPos=4'd2, setTriggerSetting=4'd3,  
setSource=4'd4, setHz=4'd5, setAmpVal=4'd6;  
  
//is a blank b/c no one ever uses ascii 255  
parameter blank=255;  
  
//when anything changes  
always @ (secPerDiv or triggerVal or triggerPos or DDSSource or settingSelect or hzVal or ampVal or  
triggerSetting) begin  
    //if it's your state you get to pick what ascii value gets sent to each hex_char  
    //Sec per div is in base 10, everything else is base 16  
    //hexDisp7 and hexDisp7 show the setting, and the rest is the value of the setting  
    if (settingSelect == setSecPerDiv)  
        begin  
            hexDisp0 <= (secPerDiv > 2) ? 6 : 3;  
            hexDisp1 <= "-";  
            hexDisp2 <= "e";  
            hexDisp3 <= (secPerDiv == 2 || secPerDiv == 5) ? 1 : 0;  
            hexDisp4 <= (secPerDiv == 1 || secPerDiv == 4) ? 1 : 0;  
            hexDisp5 <= (secPerDiv == 0 || secPerDiv == 3) ? 1 : 0;  
            hexDisp6 <= "d";  
            hexDisp7 <= "s";  
        end  
    else if (settingSelect == setTriggerVal)  
        begin  
            hexDisp0 <= triggerVal[3:0];  
            hexDisp1 <= triggerVal[7:4];  
            hexDisp2 <= blank;  
            hexDisp3 <= "h";  
            hexDisp4 <= blank;  
            hexDisp5 <= blank;  
            hexDisp6 <= "x";  
            hexDisp7 <= "t";  
        end  
    else if (settingSelect == setTriggerPos)  
        begin  
            hexDisp0 <= triggerPos[3:0];  
            hexDisp1 <= triggerPos[7:4];  
            hexDisp2 <= {2'b0,triggerPos[9:8]};  
            hexDisp3 <= "h";  
            hexDisp4 <= blank;  
            hexDisp5 <= blank;  
            hexDisp6 <= "p";  
            hexDisp7 <= "t";  
        end  
    else if (settingSelect == setTriggerSetting)  
        begin  
            hexDisp0 <= (triggerSetting == 0) ? "o" : "e";  
            hexDisp1 <= (triggerSetting == 0) ? "t" : "c";  
            hexDisp2 <= (triggerSetting == 0) ? "u" : "n";  
            hexDisp3 <= (triggerSetting == 0) ? "A" : "o";  
            hexDisp4 <= blank;  
            hexDisp5 <= blank;  
        end  
end
```

```

        hexDisp6 <= "s";
        hexDisp7 <= "t";
    end
    else if (settingSelect == setSource)
    begin
        hexDisp0 <= (DDSSource == 0) ? "e" : (DDSSource == 1) ? "r" : blank;
        hexDisp1 <= (DDSSource == 0) ? "n" : (DDSSource == 1) ? "u" : "i";
        hexDisp2 <= (DDSSource == 0) ? "i" : (DDSSource == 1) ? "q" : "r";
        hexDisp3 <= (DDSSource == 0) ? "s" : (DDSSource == 1) ? "s" : "t";
        hexDisp4 <= blank;
        hexDisp5 <= blank;
        hexDisp6 <= "s";
        hexDisp7 <= "s";
    end
    else if (settingSelect == setHz)
    begin
        hexDisp0 <= hzVal[3:0];
        hexDisp1 <= hzVal[7:4];
        hexDisp2 <= hzVal[11:8];
        hexDisp3 <= hzVal[15:12];
        hexDisp4 <= hzVal[19:16];
        hexDisp5 <= hzVal[23:20];
        hexDisp6 <= "z";
        hexDisp7 <= "h";
    end
    else if (settingSelect == setAmpVal)
    begin
        hexDisp0 <= (1 << ampVal[3:0]);
        hexDisp1 <= "/";
        hexDisp2 <= 1;
        hexDisp3 <= blank;
        hexDisp4 <= blank;
        hexDisp5 <= blank;
        hexDisp6 <= "A";
        hexDisp7 <= "s";
    end
end
endmodule

```

### ***hex\_char.v***

```

module hex_char(
                                output reg [6:0] hexDisp,
                                input [7:0] value
                                );

parameter blank=255;

//every time the input changes, light the proper seven-segment segments
always @ (value)
begin
    case (value)
        7'd0:    hexDisp <= 7'b100_0000;
        7'd1:    hexDisp <= 7'b111_1001;
        7'd2:    hexDisp <= 7'b010_0100;
        7'd3:    hexDisp <= 7'b011_0000;
        7'd4:    hexDisp <= 7'b001_1001;
        7'd5:    hexDisp <= 7'b001_0010;
        7'd6:    hexDisp <= 7'b000_0010;
        7'd7:    hexDisp <= 7'b111_1000;
        7'd8:    hexDisp <= 7'b000_0000;
        7'd9:    hexDisp <= 7'b001_0000;
        7'hA:    hexDisp <= 7'b000_1000;
        7'hB:
    endcase
end

```

```

        hexDisp <= 7'b000_0011;
7'hC:    hexDisp <= 7'b100_0110;
7'hD:    hexDisp <= 7'b010_0001;
7'hE:    hexDisp <= 7'b000_0110;
7'hF:    hexDisp <= 7'b000_1110;
blank:   hexDisp <= 7'b111_1111;
"s":     hexDisp <= 7'b001_0010;
"A":     hexDisp <= 7'b000_1000;
"i":     hexDisp <= 7'b111_1011;
"o":     hexDisp <= 7'b010_0011;
"t":     hexDisp <= 7'b000_0111;
"m":     hexDisp <= 7'b000_0110;
"n":     hexDisp <= 7'b010_1011;
"c":     hexDisp <= 7'b010_0111;
"-":     hexDisp <= 7'b011_1111;
"e":     hexDisp <= 7'b000_0110;
"h":     hexDisp <= 7'b000_1001;
"z":     hexDisp <= 7'b010_0100;
"d":     hexDisp <= 7'b010_0001;
"q":     hexDisp <= 7'b001_1000;
"/":     hexDisp <= 7'b010_1101;
"r":     hexDisp <= 7'b010_1111;
"p":     hexDisp <= 7'b000_1100;
"u":     hexDisp <= 7'b110_0011;

default: hexDisp <= 7'b111_1111;
    endcase
end
endmodule

```