The Diagonal Vector Gaussian Finite State MAC with Cooperative Encoders and Delayed CSI

Ziv Goldfeld, Haim H. Permuter, Benjamin M. Zaidel

Ben Gurion University

November, 2012

Motivation

- Motivation
- Conferencing channel model

- Motivation
- Conferencing channel model
- Common message channel model

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result
- Achievability outline

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result
- Achievability outline
- Conferencing main result

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result
- Achievability outline
- Conferencing main result
- Vector diagonal Gaussian model

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result
- Achievability outline
- Conferencing main result
- Vector diagonal Gaussian model
- Example: Gilbert-Elliot MAC

- Motivation
- Conferencing channel model
- Common message channel model
- Common message main result
- Achievability outline
- Conferencing main result
- Vector diagonal Gaussian model
- Example: Gilbert-Elliot MAC
- Summary

Motivation (delayed state information)

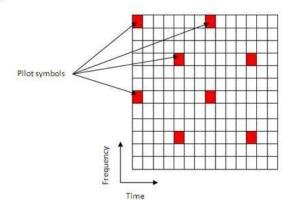
 Channel state models fading, noise and interference of uncontrolled signals.

Motivation (delayed state information)

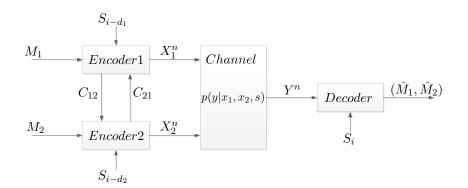
- Channel state models fading, noise and interference of uncontrolled signals.
- Channel state information (CSI) needs to be estimated.

Motivation (delayed state information)

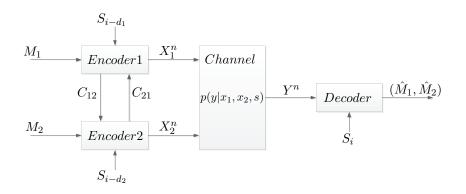
- Channel state models fading, noise and interference of uncontrolled signals.
- Channel state information (CSI) needs to be estimated.
- In LTE uplink standard, pilot signal are sent by the users in



FSM-MAC with Conferencing and Delayed CSI

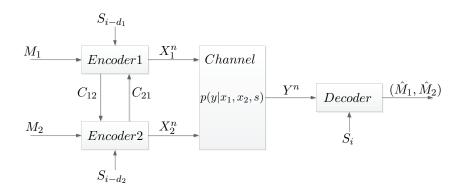


FSM-MAC with Conferencing and Delayed CSI



 CSI known to the Decoder and delayed CSI known to the Encoders.

FSM-MAC with Conferencing and Delayed CSI



- CSI known to the Decoder and delayed CSI known to the Encoders.
- Conferencing between the Encoders is possible through limited links.

• Finite number of states $|S| < \infty$.

- Finite number of states $|S| < \infty$.
- Channel state is a stationary Markov process independent of the messages.

- Finite number of states $|S| < \infty$.
- Channel state is a stationary Markov process independent of the messages.
- The random variables S_i S_{i-d} denote the channel state at time i, and i-d, respectively.

- Finite number of states $|S| < \infty$.
- Channel state is a stationary Markov process independent of the messages.
- The random variables S_i S_{i-d} denote the channel state at time i, and i-d, respectively.
- The (S_i, S_{i-d}) joint distribution is stationary and is given by

$$P(S_i = s_l, S_{i-d} = s_j) = \pi(s_j)K^d(s_l, s_j).$$

Channel Model - Partial Cooperation [Willems82]

 The conferencing takes place prior to the transmission throughout the channel.

Channel Model - Partial Cooperation [Willems82]

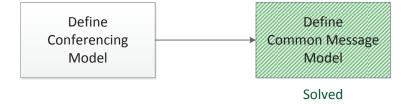
- The conferencing takes place prior to the transmission throughout the channel.
- The state process is independent of the conference communications.

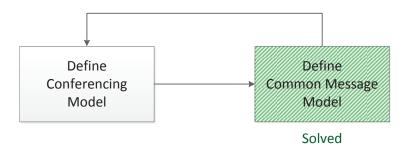
Channel Model - Partial Cooperation [Willems82]

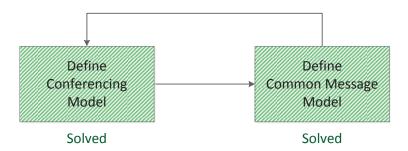
- The conferencing takes place prior to the transmission throughout the channel.
- The state process is independent of the conference communications.
- The conference is held using two communication links with finite capacities C_{12} and C_{21} .

Define Conferencing Model

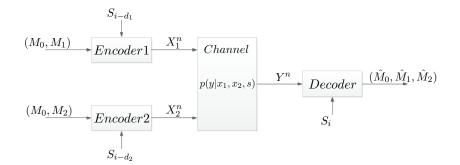








Common Message Model



Main Results Common Message with Delayed CSI $(d_1 > d_2)$

Theorem

The capacity region of FSM-MAC with a common message, CSI at the decoder and delayed CSI at the encoders with delays d_1 and d_2 , is

$$R_{1} < I(X_{1}; Y | X_{2}, U, S, \widetilde{S}_{1}, \widetilde{S}_{2}),$$

$$R_{2} < I(X_{2}; Y | X_{1}, U, S, \widetilde{S}_{1}, \widetilde{S}_{2}),$$

$$R_{1} + R_{2} < I(X_{1}, X_{2}; Y | U, S, \widetilde{S}_{1}, \widetilde{S}_{2}),$$

$$R_{0} + R_{1} + R_{2} < I(X_{1}, X_{2}; Y | S, \widetilde{S}_{1}, \widetilde{S}_{2}),$$

for some joint distribution of the form:

$$P(u|\tilde{s}_1)P(x_1|\tilde{s}_1,u)P(x_2|\tilde{s}_1,\tilde{s}_2,u).$$

The joint distribution $(S, \widetilde{S}_1, \widetilde{S}_2)$ is the same joint distribution as $(S_i, S_{i-d_1}, S_{i-d_2}).$

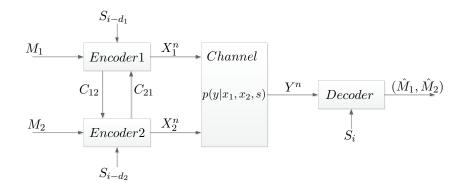
 Coding scheme: Encode using MUX, decode simultaneously using joint-typicality.

- Coding scheme: Encode using MUX, decode simultaneously using joint-typicality.
 - Achieves every possible point in the region.

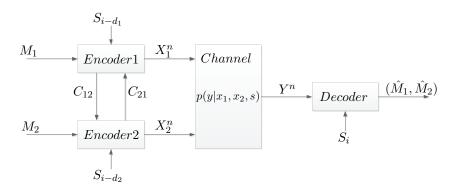
- Coding scheme: Encode using MUX, decode simultaneously using joint-typicality.
 - Achieves every possible point in the region.
 - Can be easily extended to multiple users.

- Coding scheme: Encode using MUX, decode simultaneously using joint-typicality.
 - Achieves every possible point in the region.
 - Can be easily extended to multiple users.
- Generalizes the result for the FSM-MAC with delayed CSI and no common message [Basher/Shirazy/P.11].

MAC with Conferencing and Delayed CSI

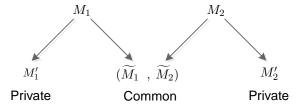


MAC with Conferencing and Delayed CSI

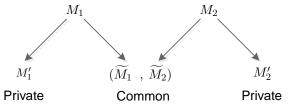


Share as much as possible of the massages through the conferencing links.

• Split the original messages (M_1,M_2) into private messages (M_1',M_2') and a common message $(\widetilde{M}_1,\widetilde{M}_2)$.

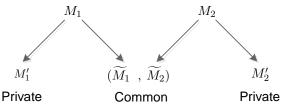


• Split the original messages (M_1,M_2) into private messages (M_1',M_2') and a common message $(\widetilde{M}_1,\widetilde{M}_2)$.



• Use the communication links in order to share $(\widetilde{M}_1, \widetilde{M}_2)$.

• Split the original messages (M_1,M_2) into private messages (M_1',M_2') and a common message $(\widetilde{M}_1,\widetilde{M}_2)$.



• Use the communication links in order to share $(\widetilde{M}_1,\widetilde{M}_2)$.

Message	Rate
$M_0' = (\widetilde{M}_1, \widetilde{M}_2)$	$C_{12} + C_{21}$
M_1'	$R_1 - C_{12}$
M_2'	$R_2 - C_{21}$

Using common message result:

Using common message result:

$$\begin{split} &(R_1-C_{12}) \leq I(X_1;Y|X_2,U,S,\tilde{S}_1,\tilde{S}_2),\\ &(R_2-C_{21}) \leq I(X_2;Y|X_1,U,S,\tilde{S}_1,\tilde{S}_2),\\ &(R_1-C_{12}) + (R_2-C_{21}) \leq I(X_1,X_2;Y|U,S,\tilde{S}_1,\tilde{S}_2),\\ &(C_{12}+C_{21}) + (R_1-C_{12}) + (R_2-C_{21}) \leq I(X_1,X_2;Y|S,\tilde{S}_1,\tilde{S}_2). \end{split}$$

Main Results with Conferencing and Delayed CSI $(d_1 \ge d_2)$

Theorem

The capacity region of FSM-MAC with partially cooperative encoders, CSI at the decoder and CSI at the encoders with delays d_1 and d_2 , is

$$R_{1} < I(X_{1}; Y | X_{2}, U, S, \tilde{S}_{1}, \tilde{S}_{2}) + C_{12},$$

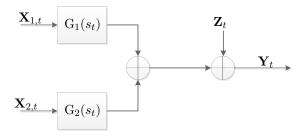
$$R_{2} < I(X_{2}; Y | X_{1}, U, S, \tilde{S}_{1}, \tilde{S}_{2}) + C_{21},$$

$$R_{1} + R_{2} < \min \left\{ \begin{array}{c} I(X_{1}, X_{2}; Y | U, S, \tilde{S}_{1}, \tilde{S}_{2}) + C_{12} + C_{21}, \\ I(X_{1}, X_{2}; Y | S, \tilde{S}_{1}, \tilde{S}_{2}) \end{array} \right\},$$

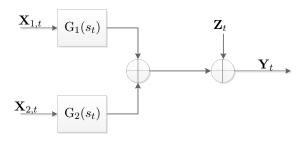
for some joint distribution of the form:

$$P(u|\tilde{s}_1)P(x_1|\tilde{s}_1,u)P(x_2|\tilde{s}_1,\tilde{s}_2,u).$$

The vector diagonal additive Gaussian noise (AGN) FSM-MAC with partially cooperative encoders and delayed CSI,



The vector diagonal additive Gaussian noise (AGN) FSM-MAC with partially cooperative encoders and delayed CSI,



The channel model,

$$\mathbf{Y}_t = \mathbf{G}_1(s_t)\mathbf{X}_{1,t} + \mathbf{G}_2(s_t)\mathbf{X}_{2,t} + \mathbf{Z}_t,$$

• $\{G_1(s)\}_{s\in\mathcal{S}}$ and $\{G_2(s)\}_{s\in\mathcal{S}}$ are real diagonal channel transition matrices of dimension $N \times N$.

- $\{G_1(s)\}_{s\in\mathcal{S}}$ and $\{G_2(s)\}_{s\in\mathcal{S}}$ are real diagonal channel transition matrices of dimension $N\times N$.
- ${\bf Z}$ is an AWGN distributed according to ${\bf Z} \sim {\cal N}(0,I)$.

- $\{G_1(s)\}_{s \in \mathcal{S}}$ and $\{G_2(s)\}_{s \in \mathcal{S}}$ are real diagonal channel transition matrices of dimension $N \times N$.
- ${\bf Z}$ is an AWGN distributed according to ${\bf Z} \sim \mathcal{N}(0, I)$.
- ullet **Z** is independent of \mathbf{X}_1 and \mathbf{X}_2 .

- $\{G_1(s)\}_{s\in\mathcal{S}}$ and $\{G_2(s)\}_{s\in\mathcal{S}}$ are real diagonal channel transition matrices of dimension $N\times N$.
- ullet ${f Z}$ is an AWGN distributed according to ${f Z} \sim \mathcal{N}(0,I).$
- \mathbf{Z} is independent of \mathbf{X}_1 and \mathbf{X}_2 .
- All vectors are real and of dimension $N \times 1$.

- $\{G_1(s)\}_{s\in\mathcal{S}}$ and $\{G_2(s)\}_{s\in\mathcal{S}}$ are real diagonal channel transition matrices of dimension $N\times N$.
- ${\bf Z}$ is an AWGN distributed according to ${\bf Z} \sim \mathcal{N}(0, I)$.
- \mathbf{Z} is independent of \mathbf{X}_1 and \mathbf{X}_2 .
- All vectors are real and of dimension $N \times 1$.
- The inputs are bounded by the following power constraints,

$$\operatorname{tr}(\Sigma_{X_1X_1}) \leq \mathcal{P}_1 \; ; \; \operatorname{tr}(\Sigma_{X_2X_2}) \leq \mathcal{P}_2.$$

Vector Diagonal Gaussian FSM-MAC - Proof Outline

• The main difficulty is to that a Gaussian triplet $(\mathbf{X}_1, \mathbf{U}, \mathbf{X}_2)$ satisfying

$$\mathbf{U} - \widetilde{S}_1 - (S, \widetilde{S}_2),$$

$$\mathbf{X}_1 - (\mathbf{U}, \widetilde{S}_1) - (S, \widetilde{S}_2),$$

$$\mathbf{X}_2 - (\mathbf{U}, \widetilde{S}_1, \widetilde{S}_2) - (\mathbf{X}_1, S),$$

is optimal.

Vector Diagonal Gaussian FSM-MAC - Proof Outline

• The main difficulty is to that a Gaussian triplet (X_1, U, X_2) satisfying

$$\begin{split} \mathbf{U} &- \widetilde{S}_1 - (S, \widetilde{S}_2), \\ \mathbf{X}_1 &- (\mathbf{U}, \widetilde{S}_1) - (S, \widetilde{S}_2), \\ \mathbf{X}_2 &- (\mathbf{U}, \widetilde{S}_1, \widetilde{S}_2) - (\mathbf{X}_1, S), \end{split}$$

is optimal.

 Use an extension of the idea of [Lapidoth/Bross/Wigger08] and [Lapidoth/Venkatesan07].

Vector Diagonal Gaussian FSM-MAC - Main Result

$$R_{1} < \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1} - d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \sum_{i=1}^{N} \log \left(1 + \left(g_{1}^{i}(s)\right)^{2} \gamma_{1}^{i}(\tilde{s}_{1})\right) + C_{12},$$

$$R_{2} < \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1} - d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \sum_{i=1}^{N} \log \left(1 + \left(g_{2}^{i}(s)\right)^{2} \gamma_{2}^{i}(\tilde{s}_{1}, \tilde{s}_{2})\right) + C_{21},$$

$$R_1 + R_2 < \frac{1}{2} \sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{\tilde{s}_2} K^{d_1 - d_2}(\tilde{s}_2, \tilde{s}_1) \sum_{s} K^{d_2}(s, \tilde{s}_2) \sum_{i=1}^{N} \log\left(1 + \left(g_1^i(s)\right)^2 \gamma_1^i(\tilde{s}_1) + \left(g_2^i(s)\right)^2 \gamma_2^i(\tilde{s}_1, \tilde{s}_2)\right) + C_{12} + C_{21},$$

$$\begin{split} R_1 + R_2 &< \frac{1}{2} \sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{\tilde{s}_2} K^{d_1 - d_2}(\tilde{s}_2, \tilde{s}_1) \sum_{s} K^{d_2}(s, \tilde{s}_2) \sum_{i=1}^N \log \Big(1 + \Big(g_1^i(s) \Big)^2 P_1^i(\tilde{s}_1) \\ &+ \big(g_2^i(s) \big)^2 P_2^i(\tilde{s}_1, \tilde{s}_2) + 2 g_1^i(s) g_2^i(s) \sqrt{\big(P_1^i(\tilde{s}_1) - \gamma_1^i(\tilde{s}_1) \big) \Big(P_2^i(\tilde{s}_1, \tilde{s}_2) - \gamma_2^i(\tilde{s}_1, \tilde{s}_2) \Big)} \Big), \end{split}$$
 subject to the constraints

subject to the constraints,

$$\sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{i=1}^{N} P_1^i(\tilde{s}_1) \leq \mathcal{P}_1, \quad ; \quad \sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{\tilde{s}_2} K^{d_1 - d_2}(\tilde{s}_2, \tilde{s}_1) \sum_{i=1}^{N} P_2^i(\tilde{s}_1, \tilde{s}_2) \leq \mathcal{P}_2,$$

$$0 \le \gamma_1^i(\tilde{s}_1) \le P_1^i(\tilde{s}_1), \forall i \in \{1, \dots, N\}, \tilde{s}_1 \in \mathcal{S},$$

$$0 \le \gamma_2^i(\tilde{s}_1, \tilde{s}_2) \le P_2^i(\tilde{s}_1, \tilde{s}_2), \forall i \in \{1, \dots, N\}, (\tilde{s}_1, \tilde{s}_2) \in \mathcal{S}^2.$$

Example: Gilbert-Elliot Gaussian MAC

- At any given time t the channel is in one of two possible states. Good or Bad.
- $g_1(G) > g_1(B)$ and $g_2(G) > g_2(B)$.

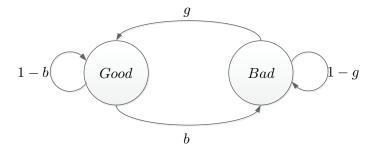
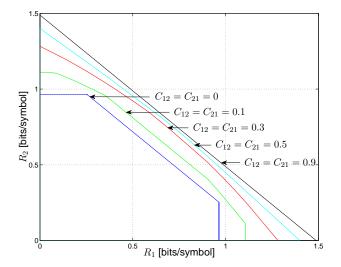


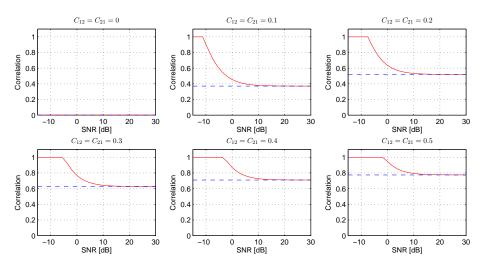
Figure: Two-state AGN channel.

Capacity region of Two-State AGN MAC Example

Fixed delays $d_1 = d_2 = 2$ and symmetrical con. $C_{12} = C_{21}$



Correlation versus SNR



 A single-letter characterization of MAC with delayed state and conferencing.

- A single-letter characterization of MAC with delayed state and conferencing.
- Conferencing: share parts of the messages.

- A single-letter characterization of MAC with delayed state and conferencing.
- Conferencing: share parts of the messages.
- Delayed state: use MUX at the encoder, joint-typicality at the decoder.

- A single-letter characterization of MAC with delayed state and conferencing.
- Conferencing: share parts of the messages.
- Delayed state: use MUX at the encoder, joint-typicality at the decoder.
- Diagonal vector Gaussian case:
 - Joint Gaussian achieve the maximum.
 - Transformed into a convex optimization problem.

- A single-letter characterization of MAC with delayed state and conferencing.
- Conferencing: share parts of the messages.
- Delayed state: use MUX at the encoder, joint-typicality at the decoder.
- Diagonal vector Gaussian case:
 - Joint Gaussian achieve the maximum.
 - Transformed into a convex optimization problem.
- Insight: Correlation is crucial in low SNR.

- A single-letter characterization of MAC with delayed state and conferencing.
- Conferencing: share parts of the messages.
- Delayed state: use MUX at the encoder, joint-typicality at the decoder.
- Diagonal vector Gaussian case:
 - Joint Gaussian achieve the maximum.
 - Transformed into a convex optimization problem.
- Insight: Correlation is crucial in low SNR.

Thank you!

Conferencing Model - Code Description

For each TX the encoding functions:

Conferencing Model - Code Description

For each TX the encoding functions:

Conferencing encoder,

$$V_{1,i} = h_{1,i}(M_1, V_2^{i-1}),$$

 $V_{2,i} = h_{2,i}(M_2, V_1^{i-1}).$

Conferencing Model - Code Description

For each TX the encoding functions:

Conferencing encoder,

$$V_{1,i} = h_{1,i}(M_1, V_2^{i-1}),$$

 $V_{2,i} = h_{2,i}(M_2, V_1^{i-1}).$

Channel encoder,

$$X_{1,i} = \left\{ \begin{array}{cc} f_{1,i}(M_1, V_2^{\ell}), & 1 \leq i \leq d_1 \\ f_{1,i}(M_1, V_2^{\ell}, S^{i-d_1}), & d_1 + 1 \leq i \leq n \end{array} \right\},\,$$

$$X_{2,i} = \left\{ \begin{array}{cc} f_{2,i}(M_2,V_1^\ell), & 1 \leq i \leq d_2 \\ f_{2,i}(M_2,V_1^\ell,S^{i-d_2}), & d_2+1 \leq i \leq n \end{array} \right\}.$$

 If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.
- Solution: Can be solved by working on the corner points and using successive decoding. [Basher/Shirazy/P.11]

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.
- Solution: Can be solved by working on the corner points and using successive decoding. [Basher/Shirazy/P.11]
- <u>Problem 2</u>: Common message generates many corner-points.

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.
- Solution: Can be solved by working on the corner points and using successive decoding. [Basher/Shirazy/P.11]
- <u>Problem 2</u>: Common message generates many corner-points.
- <u>Solution</u>: Encode using MUX, decode simultaneously using joint-typicality.

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.
- Solution: Can be solved by working on the corner points and using successive decoding. [Basher/Shirazy/P.11]
- <u>Problem 2</u>: Common message generates many corner-points.
- <u>Solution</u>: Encode using MUX, decode simultaneously using joint-typicality.
- Error analysis yield many inequalities.

- If both, the encoder and decoder, know the state (with or without delay) one can use MUX-DEMUX scheme.
 [Goldsmith/varaiya97] [Viswanathan99]
- Problem 1: Here there is an asymmetry between the encoders and the decoder.
- Solution: Can be solved by working on the corner points and using successive decoding. [Basher/Shirazy/P.11]
- <u>Problem 2</u>: Common message generates many corner-points.
- <u>Solution</u>: Encode using MUX, decode simultaneously using joint-typicality.
- Error analysis yield many inequalities.
- The inequalities are reduced using induction and the Fourier-Motzkin elimination.

• The common message M_0 is encoded only using only the "weaker" state, namely \widetilde{S}_1 .

- The common message M_0 is encoded only using only the "weaker" state, namely \widetilde{S}_1 .
- ullet The private message M_1 is encoded using \widetilde{S}_1 as well.

- The common message M_0 is encoded only using only the "weaker" state, namely \widetilde{S}_1 .
- ullet The private message M_1 is encoded using \widetilde{S}_1 as well.
- We need to split M_2 into many sub-messages according to both $(\widetilde{S}_1, \widetilde{S}_2)$. Error analysis yield many inequalities.

- The common message M_0 is encoded only using only the "weaker" state, namely \widetilde{S}_1 .
- The private message M_1 is encoded using \widetilde{S}_1 as well.
- We need to split M_2 into many sub-messages according to both $(\widetilde{S}_1, \widetilde{S}_2)$. Error analysis yield many inequalities.
- The reduction of the inequalities is proved using induction and the Fourier-Motzkin elimination.

MAC with common message need one auxiliary.

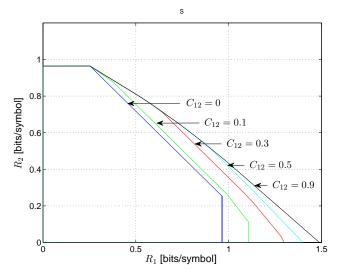
- MAC with common message need one auxiliary.
- MAC with delayed state need one auxiliary.

- MAC with common message need one auxiliary.
- MAC with delayed state need one auxiliary.
- Auxiliaries can be combined.

- MAC with common message need one auxiliary.
- MAC with delayed state need one auxiliary.
- Auxiliaries can be combined.
- Identification of the auxiliary random variable U as the common knowledge of the two encoders, $U_i = (M_0, S^{i-d_1-1})$.

Capacity region of Two-State AGN MAC Example

Fixed delays $d_1 = d_2 = 2$ and asymmetrical con. $C_{12} \ge C_{21} = 0$



Capacity region of Two-State AGN MAC Example

Fixed delays $d_1 = d_2 = 2$ and infinite con. $C_{12} \le C_{21} = \infty$

